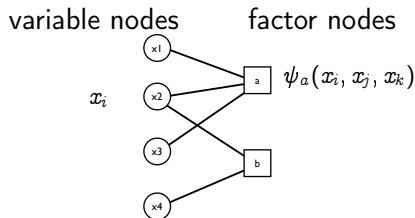


5. Density evolution

Probabilistic analysis of message passing algorithms



- consider factor graph model $G = (V, F, E)$ and

$$\mu(x) = \frac{1}{Z} \prod_{a \in F} \psi_a(x_{\partial a}) \prod_{i \in V} \psi_i(x_i)$$

- sum-product algorithm and max-product algorithms are instances of *message-passing algorithms*
 - discrete $x_i \in \mathcal{X}$
 - two sets of messages $\{\nu_{i \rightarrow a}(x_i)\}$ and $\{\tilde{\nu}_{a \rightarrow i}(x_i)\}$
 - update:

$$\nu_{i \rightarrow a}^{(t+1)} = F_{i \rightarrow a}(\{\tilde{\nu}_{b \rightarrow i}^{(t)} : b \in \partial i \setminus a\})$$

$$\tilde{\nu}_{a \rightarrow i}^{(t)} = G_{a \rightarrow i}(\{\nu_{j \rightarrow a}^{(t)} : j \in \partial a \setminus i\})$$

- assumptions for probabilistic analysis
 - ▶ a *random graph* is a graph $G = (V, F, E)$ where E is drawn randomly from a set of possible graphs
e.g., Erdős-Renyi graph, random regular graph
 - ▶ asymptotic analysis: in the limit $n \rightarrow \infty$

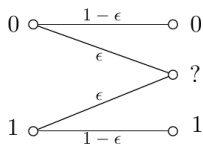
- *density evolution* is used in
 - ▶ analyzing channel codes
 - ▶ analyzing solution space of XORSAT
 - ▶ analyzing a message-passing algorithm for crowdsourcing
 - ▶ analyzing belief propagation for community detection
 - ▶ etc.

Example: channel coding

- sending messages through a noisy *channel*



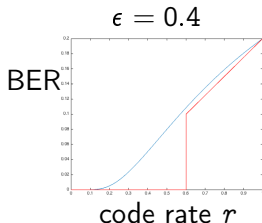
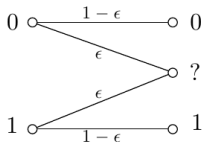
- channel is defined by $\mathbb{P}_{Y|X}(y|x)$
- Binary Erasure Channel (BEC)
 - ▶ input $x_i \in \{0, 1\}$, output $y_i \in \{0, 1, *\}$



- goal: estimate $\hat{x}_1, \dots, \hat{x}_n$ given y_1, \dots, y_n
- performance metric: average bit error probability

$$P_{\text{error}} \equiv \frac{1}{n} \sum_{i=1}^n \mathbb{P}(x_i \neq \hat{x}_i)$$

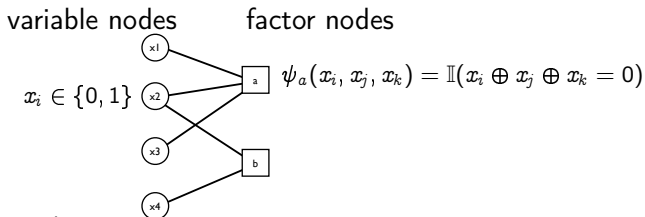
Message length vs. block length



- no coding: $(01001) \Rightarrow (01 * 0*)$
 - ▶ message $k = 5$ bits, block length $n = 5$
 \Rightarrow rate of this code $r \triangleq k/n = 1$, delay is one
 - ▶ $P_{\text{error}} = \epsilon/2$
- repetition code: $(000111000000111) \Rightarrow (0 * * 1 * 10 * 0 * * * 111)$
 - ▶ $k = 5$, $n = 15$
 - ▶ rate $r = 1/3$ and $P_{\text{error}} = \epsilon^3/2$, delay is 3
 - ▶ in general, $P_{\text{error}} = \epsilon^{1/r}/2 > 0$ (unless rate is zero)
- information theory
 - ▶ *capacity* of a BEC is $1 - \epsilon$
 - ▶ there exists a code such that $\lim_{n \rightarrow \infty} P_{\text{error}} = 0$ with rate $r < 1 - \epsilon$
 - ▶ using the BEC n times, one can reliably send $k = (1 - \epsilon)n$ bits of

Modern coding theory

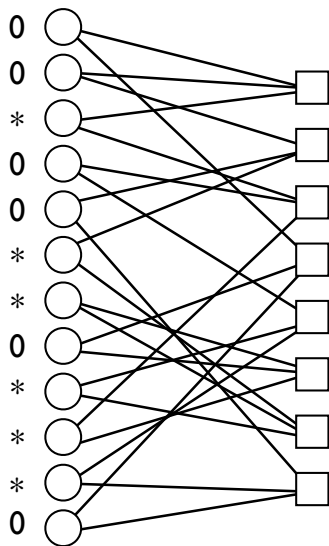
- modern codes = iterative decoding (belief propagation)
 - ▶ Turbo code
 - ▶ Low-Density Parity Check (LDPC) code
 - ▶ Polar code
 - ▶ etc.
- LDPC code is defined by a factor graph model



- ▶ block length $n = 4$
- ▶ number of factors $m = 2$
- ▶ allowed messages = $\{0000, 0111, 1010, 1101\}$
- ▶ message size $k \triangleq \log_2(\# \text{ of allowed messages}) = 2$ ($k = n - m$)
- ▶ rate $r \triangleq k/n = 1/2$
- ▶ received $y = (0 * 1*)$, then $\hat{x} = (0111)$
- ▶ received $y = (0 * **)$, then ?

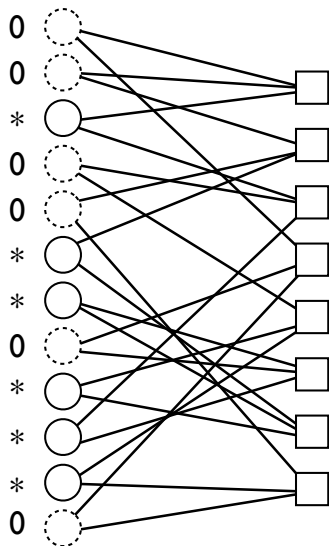
Peeling decoder is equivalent to sum-product for BEC

without loss of generality, suppose all 0's sent



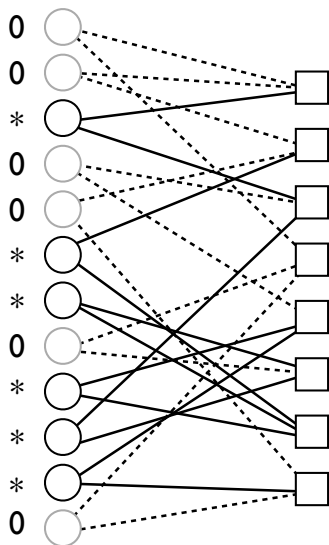
Peeling decoder is equivalent to sum-product for BEC

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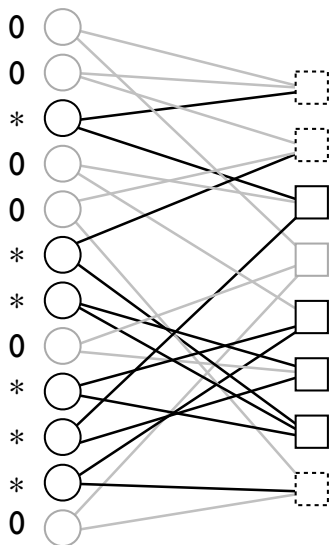
Peeling decoder is equivalent to sum-product for BEC

without loss of generality, suppose all 0's sent



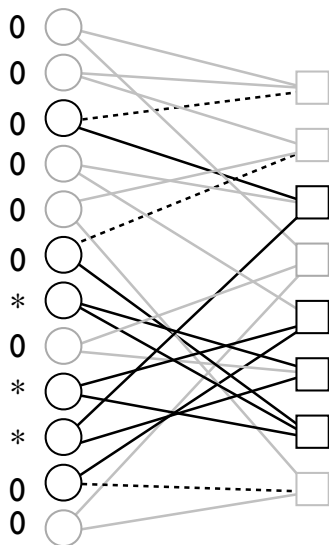
Peeling decoder is equivalent to sum-product for BEC

without loss of generality, suppose all 0's sent



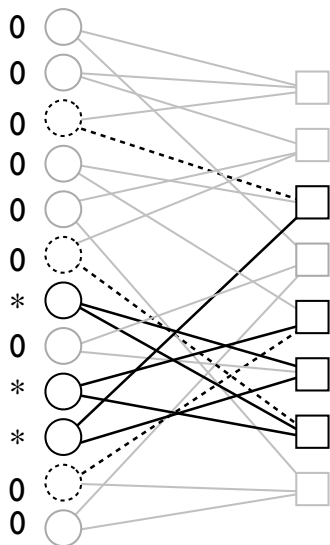
Peeling decoder is equivalent to sum-product for BEC

without loss of generality, suppose all 0's sent



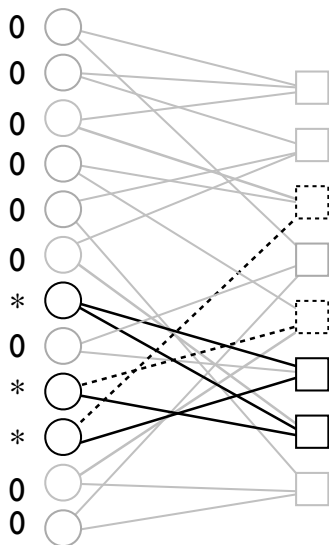
Peeling decoder is equivalent to sum-product for BEC

without loss of generality, suppose all 0's sent

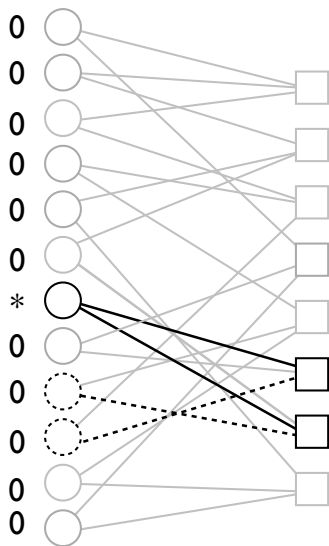


Peeling decoder is equivalent to sum-product for BEC

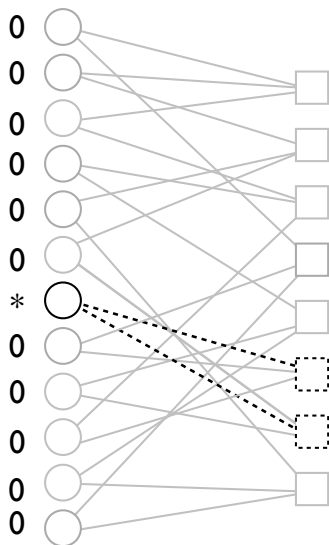
without loss of generality, suppose all 0's sent



Peeling decoder is equivalent to sum-product for BEC
without loss of generality, suppose all 0's sent



Peeling decoder is equivalent to sum-product for BEC
without loss of generality, suppose all 0's sent



Modern coding theory

- decoding using belief propagation

$$\mu_y(\mathbf{x}) = \frac{1}{Z} \prod_{i \in V} \mathbb{P}_{Y|X}(y_i | x_i) \prod_{a \in F} \mathbb{I}(\oplus x_{\partial a} = 0)$$

- use (parallel) sum-product algorithm to find $\mu(x_i)$ and let

$$\hat{x}_i = \arg \max \mu(x_i)$$

- ▶ minimizes bit error rate

Decoding by sum-product algorithm

- Directly applying parallel sum-product algorithm

$$\nu_{i \rightarrow a}^{(t+1)}(\mathbf{x}_i) = \mathbb{P}(y_i | \mathbf{x}_i) \prod_{b \in \partial i \setminus \{a\}} \tilde{\nu}_{b \rightarrow i}^{(t)}(\mathbf{x}_i)$$

$$\tilde{\nu}_{a \rightarrow i}^{(t+1)} = \sum_{x_{\partial a \setminus \{i\}}} \prod_{j \in \partial a \setminus \{i\}} \nu_{j \rightarrow a}(\mathbf{x}_j) \mathbb{I}(\oplus x_{\partial a} = 0)$$

- Notice that all $\nu, \tilde{\nu}$'s can take only one of the following three values:

$$\nu_{i \rightarrow a}(\mathbf{x}_i) \in \left\{ \begin{bmatrix} 1 \\ 0 \end{bmatrix}, \begin{bmatrix} 0 \\ 1 \end{bmatrix}, \begin{bmatrix} 1/2 \\ 1/2 \end{bmatrix} \right\}$$

hence, we will map these vectors to symbols $\{0, 1, *\}$

- because (proof by induction)
 - ▶ initially,

$$\tilde{\nu}_{a \rightarrow i}^{(0)}(\mathbf{x}_i) = \begin{bmatrix} 1/2 \\ 1/2 \end{bmatrix}, \quad \nu_{i \rightarrow a}^{(1)}(\mathbf{x}_i) = \begin{cases} \begin{bmatrix} 1 \\ 0 \end{bmatrix} & \text{if } y_i = 0 \\ \begin{bmatrix} 0 \\ 1 \end{bmatrix} & \text{if } y_i = 1 \\ \begin{bmatrix} 1/2 \\ 1/2 \end{bmatrix} & \text{if } y_i = * \end{cases}$$

- ▶ recursively, assuming the input messages up to t are one of the three types,

$$\tilde{\nu}_{a \rightarrow i}^{(t+1)}(x_i) = \begin{cases} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} & \text{if all other bits are determined and add up to 0} \\ \begin{bmatrix} 0 \\ 1 \\ 1 \end{bmatrix} & \text{if all other bits are determined and add up to 1} \\ \begin{bmatrix} 1/2 \\ 1/2 \end{bmatrix} & \text{if there is at least one bit that is not determined} \end{cases}$$

$$\nu_{i \rightarrow a}^{(t+1)}(x_i) = \begin{cases} \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} & \text{if at least one of the input message is } \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} \\ \begin{bmatrix} 0 \\ 1 \\ 1 \end{bmatrix} & \text{if at least one of the input message is } \begin{bmatrix} 0 \\ 1 \\ 1 \end{bmatrix} \\ \begin{bmatrix} 1/2 \\ 1/2 \end{bmatrix} & \text{if all input messages are } \begin{bmatrix} 1/2 \\ 1/2 \end{bmatrix} \end{cases}$$

- ▶ consequently, the messages only take those three values
- ▶ we will denote those three types of messages as 0, 1, and *, meaning determined to be 0 or 1, or not determined.

- (simplified) Parallel sum-product for BEC

- ▶ $\nu_{i \rightarrow a}^{(t)} \in \{0, 1, *\}$ our belief about x_i
- ▶ $\tilde{\nu}_{a \rightarrow i}^{(t)} \in \{0, 1, *\}$ our belief about x_i
- ▶ at iteration 0: $\nu_{i \rightarrow a}^{(0)} = y_i$
- ▶ at iteration t :

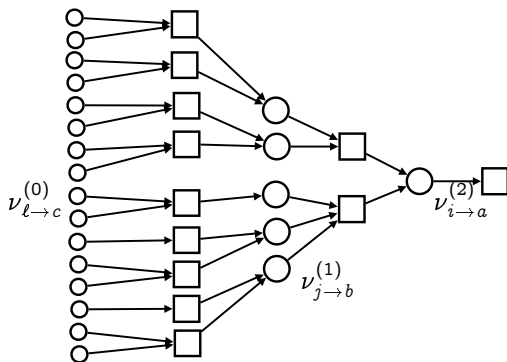
$$\tilde{\nu}_{a \rightarrow i}^{(t)} = \begin{cases} * & \text{if any of the incoming messages is a } * \\ \bigoplus x_{\partial a \setminus i} & \text{otherwise} \end{cases}$$

$$\nu_{i \rightarrow a}^{(t)} = \begin{cases} * & \text{if all of the incoming messages are } * \\ x_{b \rightarrow i} & \text{otherwise} \end{cases}$$

- this is equivalent to the peeling decoder

Probabilistic analysis: density evolution

- an LDPC code is defined by a graph G
- probabilistic analysis: we want to predict the performance of a given LDPC code G
- to this end, we use **density evolution** on the **computation tree**
- if G is **locally tree like** up to depth k , and if we run sum-product algorithm for k iterations, then the resulting message $\nu_{i \rightarrow a}^{(k)}$ is fully described by the computation tree for the message $\nu_{i \rightarrow a}^{(k)}$:



- however, it is not always possible to apply density evolution
- a few assumptions
 - ▶ sparse random graph construction
(e.g. random (ℓ, r) -regular graph from the configuration model)
 - ▶ asymptotic analysis:
in the limit $n \rightarrow \infty$ but finite number of iterations t
- why do we need these assumptions?
 - ▶ it is difficult to analyze one particular graph, so we resort to the expected performance where the expectation also take into account the randomness in the graph generation
 - ▶ random sparse graphs are locally tree-like
 - ★ if we consider random (d, d) -regular graphs, the expected number of 2-cycles is $(\frac{1}{n} + \dots + \frac{d-1}{n}) \times n$, which is small compared to the number of edges

Probabilistic analysis: density evolution

- locally-tree like structure ensures that the incoming messages are independent
- formally, as $n \rightarrow \infty$ local neighborhood of a node converges in probability to a random tree

$$\mathbb{P}\left(\lim_{n \rightarrow \infty} \text{depth } k \text{ neighborhood of a random } i \text{ is a tree}\right) = 1$$

- density evolution** for (ℓ, r) -regular graph

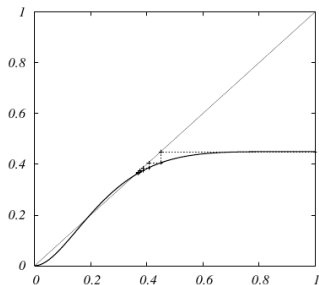
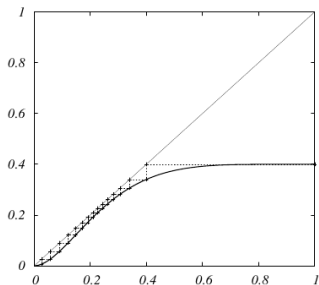
- $z_t \in [0, 1]$ be the probability a randomly chosen message from $\{\nu_{i \rightarrow a}^{(t)}\}$ is an erasure
- $w_t \in [0, 1]$ be the probability a randomly chosen message from $\{\tilde{\nu}_{a \rightarrow i}^{(t)}\}$ is an erasure
- in the limit $n \rightarrow \infty$, they satisfy the *density evolution equations*

$$\begin{aligned}w_t &= 1 - (1 - z_{t-1})^{r-1} \\z_t &= \epsilon w_t^{\ell-1}\end{aligned}$$

$$z_t = \epsilon(1 - (1 - z_{t-1})^{r-1})^{\ell-1}$$

with initial condition $z_0 = \epsilon$

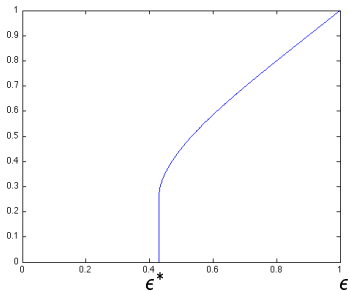
- density evolution for (3,6) code with $\epsilon = 0.4$ (left) and 0.45 (right)



- rate of this code = 0.5, threshold $\epsilon^* \simeq 0.4xxx$,
- this simple code achieves rate less than the capacity = $1 - \epsilon$
- $P_{\text{error}}(t) = \lim_{n \rightarrow \infty} P_{\text{error}}(n, t)$
- analyze $\lim_{t \rightarrow \infty} \lim_{n \rightarrow \infty} P_{\text{error}}(n, t)$, is this what we want?

for a given value of ϵ , we can **numerically** run the density evolution, since it is an evolution of a scalar value, which gives

bit error rate of (3, 6)-codes

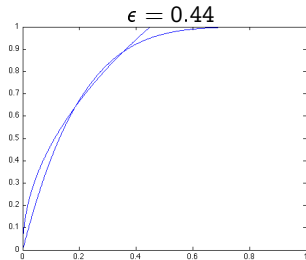
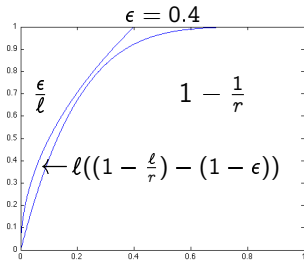


how do we find ϵ^* ?

$$z_t = \epsilon(1 - (1 - z_{t-1})^{r-1})^{\ell-1}$$

let's change the equation to

$$\left(\frac{z_t}{\epsilon}\right)^{1/(\ell-1)} = 1 - (1 - z_{t-1})^{r-1}$$



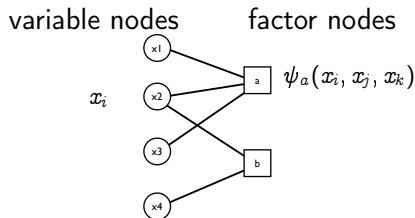
for a given ϵ , if there is no overlap, then achieve zero error probability

$$\int_0^1 \epsilon y^{\ell-1} dy = \frac{\epsilon}{\ell}, \quad \int_0^1 (1 - (1-x)^{r-1}) dx = 1 - \frac{1}{r}$$

rate of the code = $1 - \frac{\ell}{r}$ vs. capacity = $1 - \epsilon$

extend this analysis to construct capacity achieving *tornado codes*

density evolution for general message passing algorithms



- consider factor graph model $G = (V, F, E)$ and

$$\mu(x) = \frac{1}{Z} \prod_{a \in F} \psi_a(x_{\partial a}) \prod_{i \in V} \psi_i(x_i)$$

- update:
 $\nu_{i \rightarrow a}^{(t+1)} = F_{i \rightarrow a}(\{\tilde{\nu}_{b \rightarrow i}^{(t)} : b \in \partial i \setminus a\})$
 $\tilde{\nu}_{a \rightarrow i}^{(t)} = G_{a \rightarrow i}(\{\nu_{j \rightarrow a}^{(t)} : j \in \partial a \setminus i\})$
- density evolution equation

$$z^{(t+1)} = F(w_1^{(t)}, \dots, w_{\ell-1}^{(t)})$$
$$w^{(t)} = G(z_1^{(t)}, \dots, z_{k-1}^{(t)})$$

- formally, as $n \rightarrow \infty$ a randomly chosen message from $\{\nu_{i \rightarrow a}^{(t)}\}$ converge in probability to $z^{(t)}$
- who cares about random graphs?
- who cares about asymptotics?

alphabet $x_i \in \mathcal{X}$	messages $\nu_{i \rightarrow a} \in \mathcal{Y}$	density \mathcal{Z}
discrete $\{0, 1\}$	discrete $\{0, 1, *\}$	continuous \mathbb{R}
discrete	continuous $\mathbb{R}^{ \mathcal{X} -1}$	distribution over $\mathbb{R}^{ \mathcal{X} -1}$
continuous \mathbb{R}	distribution over \mathbb{R}	dist. over dist. over \mathbb{R}

- how do we compute evolution of distributions?
 - ▶ quantization
 - ▶ Gaussian approximation
 - ▶ *population dynamics*: represent the density using 'samples'