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## Introduction to Digital Data Acquisition:

# Sampling

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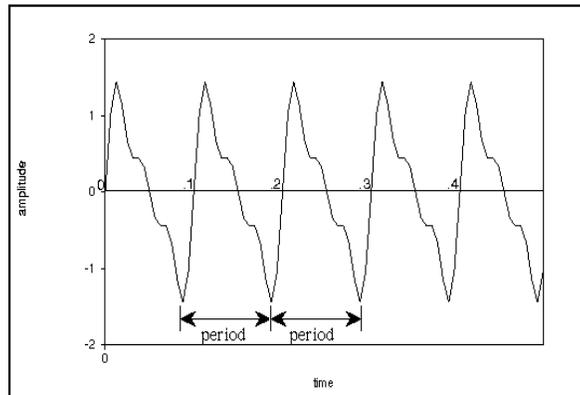
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### Physical world is analog

- Digital systems need to
    - Measure analog quantities
      - Switch inputs, speech waveforms, etc
    - Control analog systems
      - Computer monitors, automotive engine control, etc
  - Analog-to-digital: A/D converter (ADC)
    - Example: CD recording
  - Digital-to-analog: D/A converter (DAC)
    - Example: CD playback
-

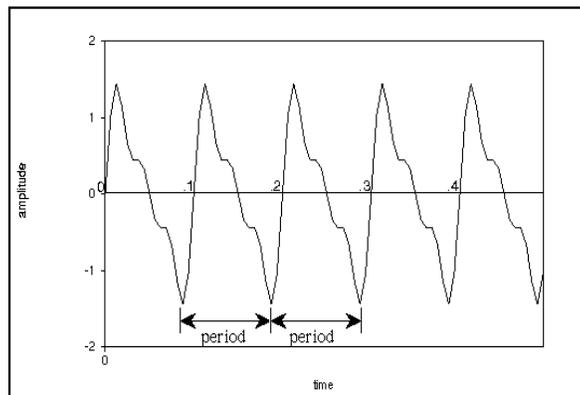
## A little background

- For periodic waveforms, the duration of the waveform before it repeats is called the period of the waveform



## Frequency

- the rate at which a regular vibration pattern repeats itself (frequency =  $1/\text{period}$ )



## Frequency of a Waveform

- The unit for frequency is cycles/second, also called Hertz (Hz).
- The frequency of a waveform is equal to the reciprocal of the period.

$$\text{frequency} = 1/\text{period}$$

## Frequency of a Waveform

- Examples:

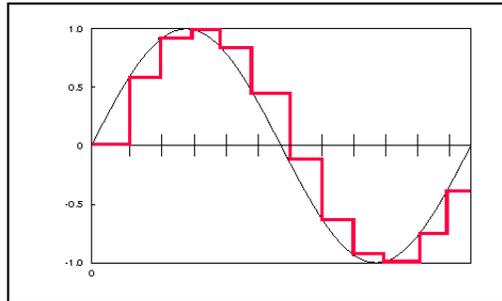
frequency = 10 Hz  
period = .1 (1/10) seconds

frequency = 100 Hz  
period = .01 (1/100) seconds

frequency = 261.6 Hz (middle C)  
period = .0038226 (1/ 261.6) seconds

## Waveform Sampling

- To represent waveforms in digital systems, we need to digitize or sample the waveform.



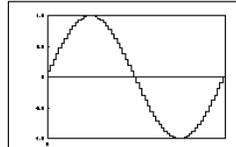
- side effects of digitization:
  - introduces some noise
  - limits the maximum upper frequency range

## Sampling Rate

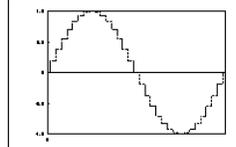
- The sampling rate (SR) is the rate at which amplitude values are digitized from the original waveform.
  - CD sampling rate (high-quality):  
SR = 44,100 samples/second
  - medium-quality sampling rate:  
SR = 22,050 samples/second
  - phone sampling rate (low-quality):  
SR = 8,192 samples/second

## Sampling Rate

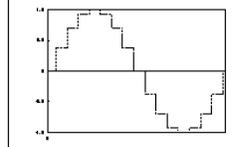
- Higher sampling rates allow the waveform to be more accurately represented



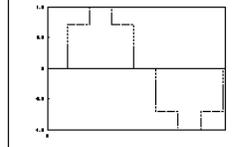
64 samples/cycle



32 samples/cycle



16 samples/cycle



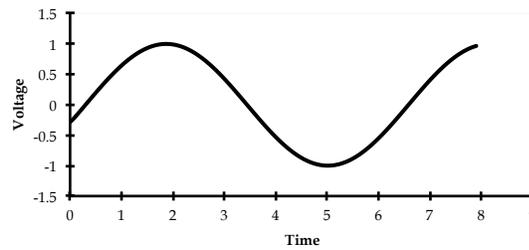
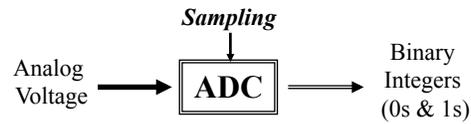
8 samples/cycle

## Digital Data Acquisition

- Data Representation - *Digital vs. Analog*
- Analog-to-Digital Conversion
- Number Systems
  - Binary Numbers
  - Binary Arithmetic
- Sampling & Aliasing

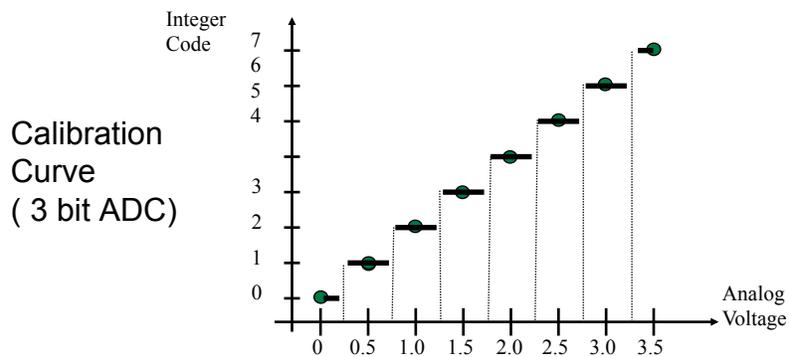
## Analog-to-Digital Conversion

- Converts analog voltages to binary integers.



## Analog-to-Digital Conversion

### • ADC calibration



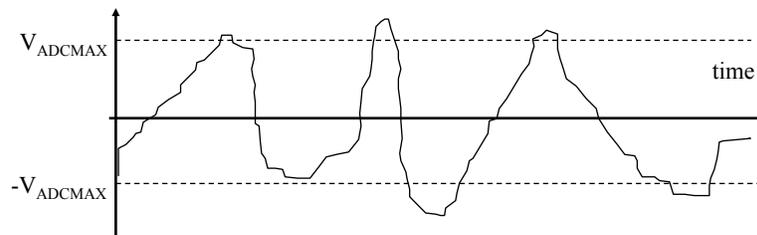
## Analog-to-Digital Conversion

### ■ Input Range

- Unipolar:  $(0, V_{ADC\text{MAX}})$
- Bipolar:  $(-V_{ADC\text{MAX}}, +V_{ADC\text{MAX}})$  (Nominal Range)

### □ Clipping:

If  $|V_{IN}| > |V_{ADC\text{MAX}}|$ , then  $|V_{OUT}| = |V_{ADC\text{MAX}}|$



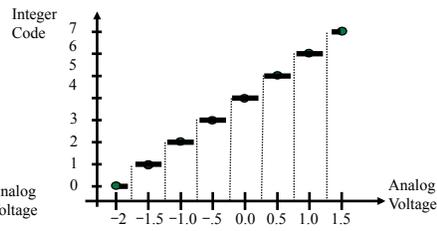
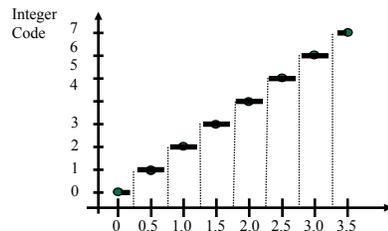
## Analog-to-Digital Conversion

### ■ Quantization Interval (Q)

- $n$  bit ADC, the input range is divided into  $2^n - 1$  intervals.

- 3 bit ADC:

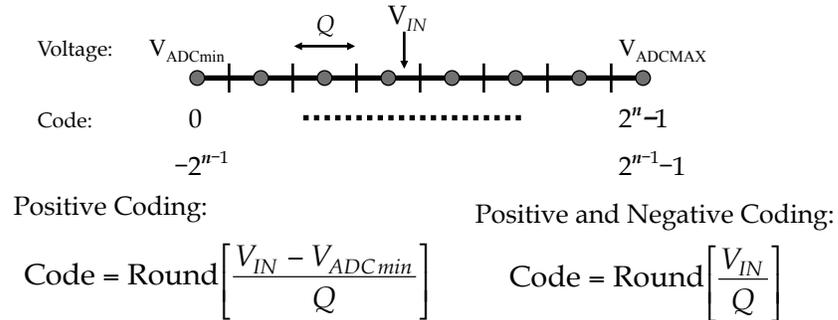
$$Q = \frac{V_{ADC\text{MAX}} - V_{ADC\text{min}}}{2^n - 1}$$



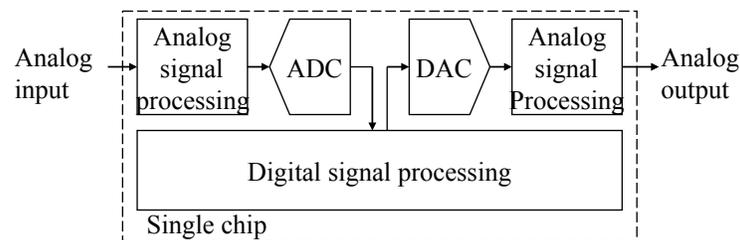
## Analog-to-Digital Conversion

### ■ Voltage to Integer Code

- $n$  bit ADC

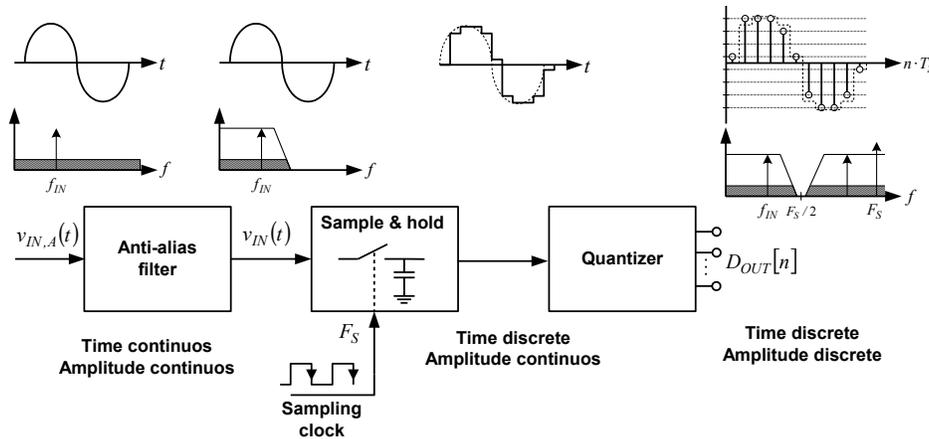


## Why A/D-conversion?



- Signals are analog by nature
- ADC necessary for DSP
- Digital signal processing provides:
  - Close to infinite SNR
  - Low system cost
  - Repetitive system
- ADC bottle necks:
  - Dynamic range
  - Conversion speed
  - Power consumption

## A/D-converter basics

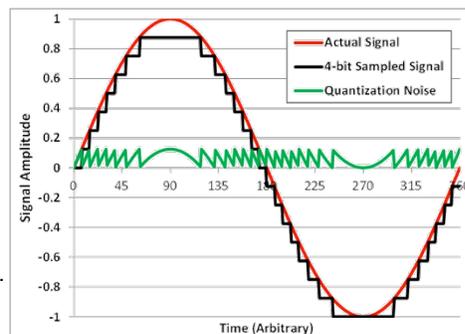


$$D_{OUT}^{ideal}[n] = G_{ideal} \cdot v_{IN}(n \cdot T_S) + q(n)$$

$$D_{OUT}^{real}[n] = G_{ideal} \cdot (1 + \varepsilon) \cdot v_{IN}(n \cdot T_S) + q(n) + e_{offset}(n) + e_{noise}(n) + e_{jitter}(n) + e_{distortion}(n)$$

## The Theory

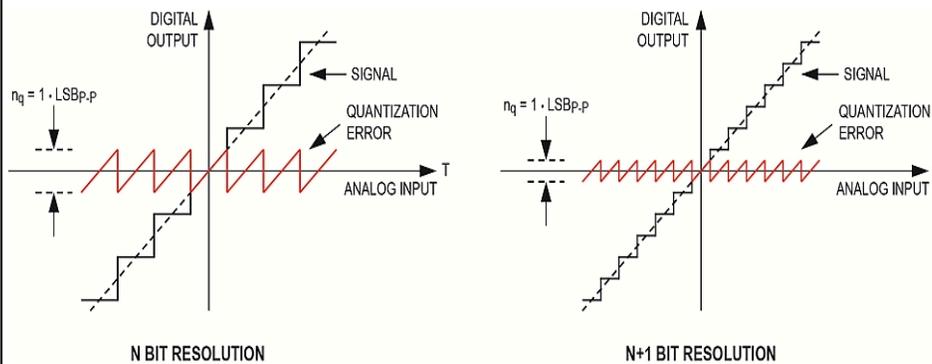
- Sampling theory is a subset of communications theory
  - Same basic math
    - Want to record signal, not noise
  - **Quantization**: Conversion from analog to discrete values
  - **Coding**: Assigning a digital word to each discrete value
    - Thermometer code, Gray code...
- Quantization adds noise
  - Analog signal is continuous
  - Digital representation is approximate
  - Difference (error) is noiselike



## Some terminology

- Resolution (n): Number of states in bits
  - Example: A 3-bit A/D
- Full-scale range (FSR): The input or output voltage range
  - Example: ADC inputs outside the FSR are always 111 or 000
- Step size (Q):  $\frac{FSR}{2^n}$
- RMS quantization error:  $\frac{Q}{\sqrt{12}}$ 
  - RMS value of triangle wave

## Quantization noise



•N-bit converter: 
$$\delta = \frac{V_{FSR}}{2^N}$$

## Quantization noise

- Noise energy:

$$V_{Q(RMS)} = \sqrt{\frac{1}{\delta} \int_{-\delta/2}^{\delta/2} V_Q^2 dV_Q} = \sqrt{\frac{\delta^2}{12}}$$

- Signal energy:

$$V_{in(RMS)} = \frac{\delta \cdot 2^N}{2\sqrt{2}}$$

- SNR for ideal ADC:

$$SNR = 20 \log\left(\frac{V_{in(RMS)}}{V_{Q(RMS)}}\right)$$

$$SNR = 20 \log\left(2^N \cdot \sqrt{\frac{3}{2}}\right)$$

$$SNR = 6.02 \times N + 1.76 [dB]$$

## Quantization noise

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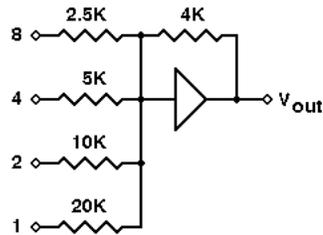
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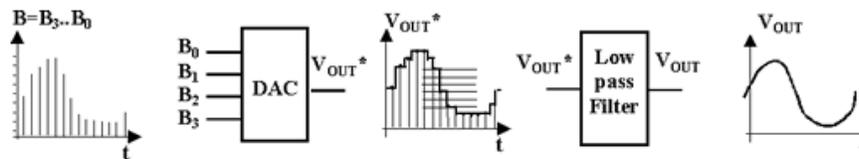
$$SNR = 6.02 \times N + 1.76 [dB]$$

## D/A converters

- Easier to design and use than A/Ds
- Types
  - Weighted current source DAC
  - R-2R DAC
  - Multiplying DAC
- Need to smooth the output

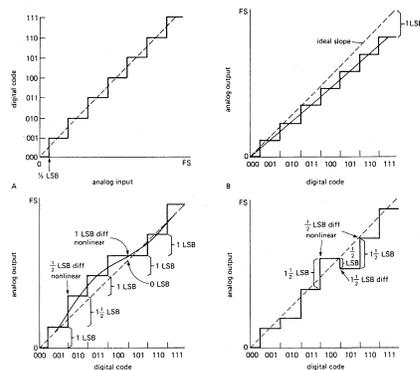


Digital to Analog Converter converts a digital signal to an analog output



## You will use DACs

- DAC specs are tricky!
  - Check the errors
  - Check the settling
- Vendors use deceptive advertising
  - 16-bit DAC!!!
    - But errors may give only 12-bit accuracy
    - You have to figure this out from the specs



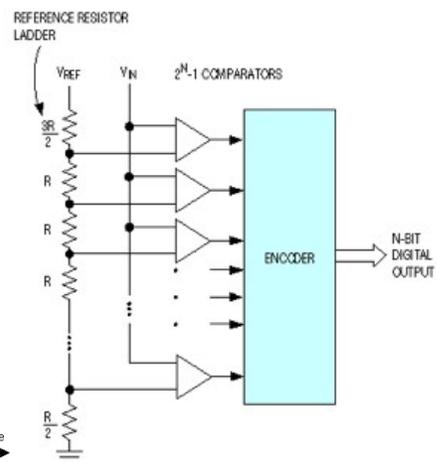
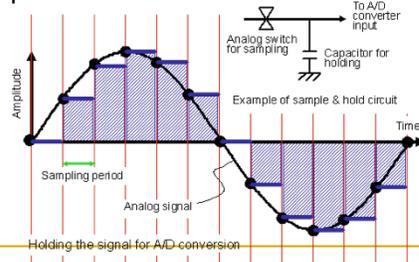
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## A/D converters

- **Hard** to design
- Contain digital parts
  - Encoders
  - FSMs
- Many types
  - Successive approximation
  - Flash
  - Pipelined-flash
  - Integrating
  - Sigma-delta
  - Charge balanced
  - Folding
  - Others

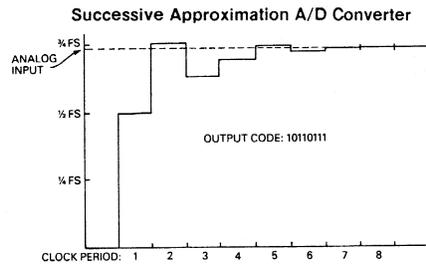
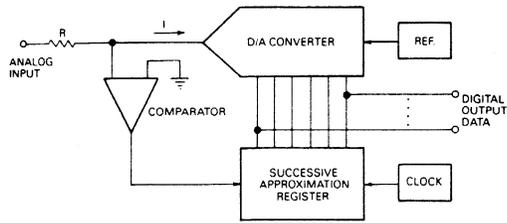
## Example: Flash A/D

- Advantages
  - Ultra-fast
- Disadvantages
  - High power
  - Low resolution
  - Metastability
- Sample/hold improves performance



## Example: Successive approximation ADC

- Advantages
  - Low power
  - High resolution
- Disadvantages
  - Slow
- Problem: DAC must settle to LSB accuracy at every step

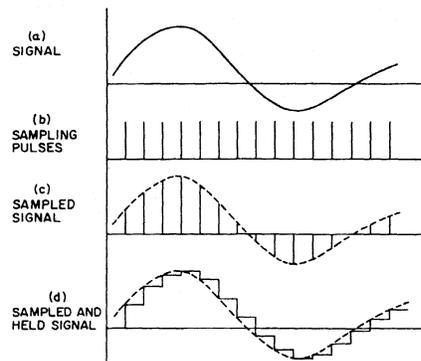


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D/A Output for 8-Bit Successive Approximation Conversion

## Sampling

- Quantizing a signal
  - 1) We sample it
  - 2) We encode the samples
- Questions:
  - How fast do we sample?
  - How do we do this in hardware?
  - What resolution do we need?



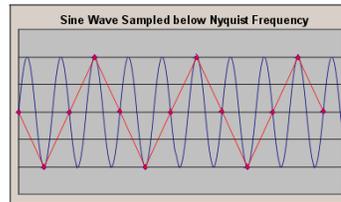
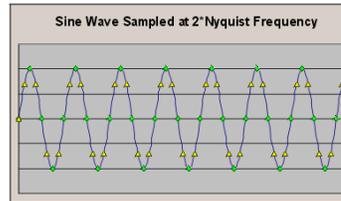
Signal Sampling

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## Shannon's sampling theorem

If a continuous, band-limited signal contains no frequency components higher than  $f_c$ , then we can recover the original signal without distortion if we sample at a rate of at least  $2f_c$  samples/second

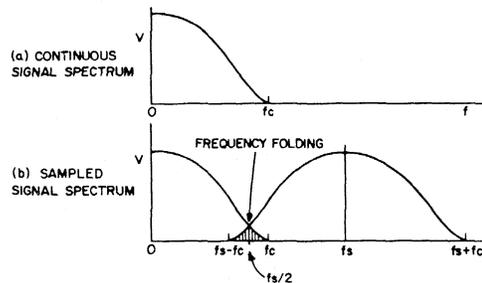
- ◆  $2f_c$  is called the Nyquist rate
- ◆ Real life
  - ⇒ Sample at  $2.5f_c$  or faster
  - ⇒ Sample clock should not be coherent with the input signal



<http://www.videomicroscopy.com/vancouverlecture/nyquist.htm>

## Frequency domain analysis

- Take the Fourier Transform of the signal
  - Shows a signal's *frequency* components
- Undersampled frequency components fold back!

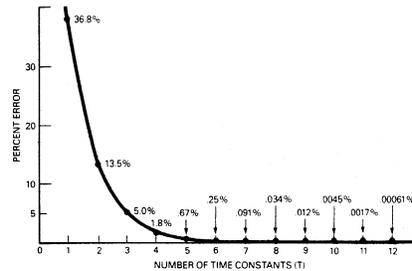


Frequency Spectra Demonstrating the Sampling Theorem

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## Sampling speed versus bit resolution

- Hardware issues
  - Sampling speed depends on bit resolution
    - Think time constants
    - Settling error =  $e^{-\frac{t}{\tau}}$
- Examples:
  - 8-bit resolution takes  $5.5\tau$
  - 12-bit resolution takes  $8.3\tau$
  - 16-bit resolution takes  $11\tau$



Output Settling Error as a Function of Number of Time Constants

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## Nyquist–Shannon sampling theorem

- A theorem, developed by Harry Nyquist, and proven by Claude Shannon, which states that an analog signal waveform may be uniquely reconstructed, without error, from samples taken at equal time intervals.

## Nyquist–Shannon sampling theorem

- The sampling rate must be equal to, or greater than, twice the highest frequency component in the analog signal.
- Stated differently:
- The highest frequency which can be accurately represented is one-half of the sampling rate.

## Nyquist Theorem and Aliasing

- Nyquist Theorem:  
We can digitally represent only frequencies up to half the sampling rate.
  - Example:  
CD: SR=44,100 Hz  
Nyquist Frequency =  $SR/2 = 22,050$  Hz
  - Example:  
SR=22,050 Hz  
Nyquist Frequency =  $SR/2 = 11,025$  Hz

## Nyquist Theorem and Aliasing

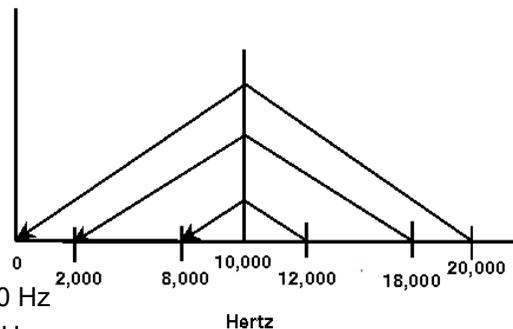
- Frequencies above Nyquist frequency "fold over" to sound like lower frequencies.
  - This foldover is called *aliasing*.
- Aliased frequency  $f$  in range  $[SR/2, SR]$  becomes  $f'$ :
$$f' = |f - SR/2|$$

## Nyquist Theorem and Aliasing

$$f' = |f - SR/2|$$

- Example:

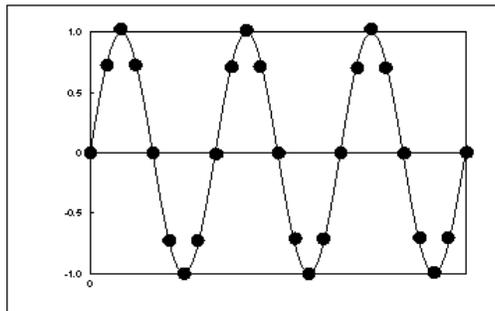
- $SR = 20,000$  Hz
- Nyquist Frequency =  $10,000$  Hz
- $f = 12,000$  Hz  $\rightarrow f' = 8,000$  Hz
- $f = 18,000$  Hz  $\rightarrow f' = 2,000$  Hz
- $f = 20,000$  Hz  $\rightarrow f' = 0$  Hz



## Nyquist Theorem and Aliasing

### ■ Graphical Example 1a:

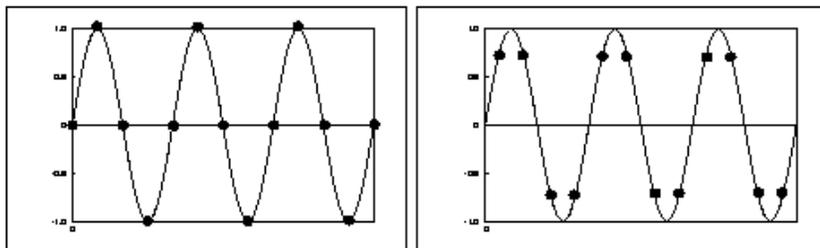
- SR = 20,000 Hz
- Nyquist Frequency = 10,000 Hz
- $f = 2,500$  Hz (no aliasing)



## Nyquist Theorem and Aliasing

### ■ Graphical Example 1b:

- SR = 20,000 Hz
- Nyquist Frequency = 10,000 Hz
- $f = 5,000$  Hz (no aliasing)

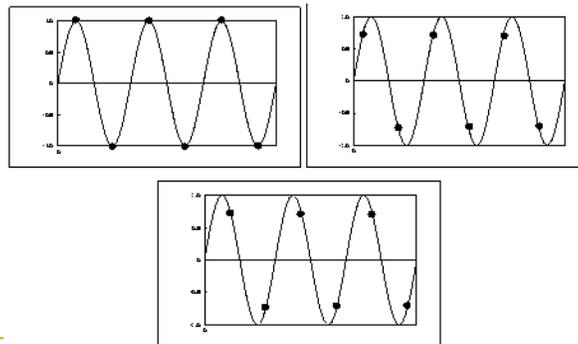


(left and right figures have same frequency, but have different sampling points)

## Nyquist Theorem and Aliasing

### ■ Graphical Example 2:

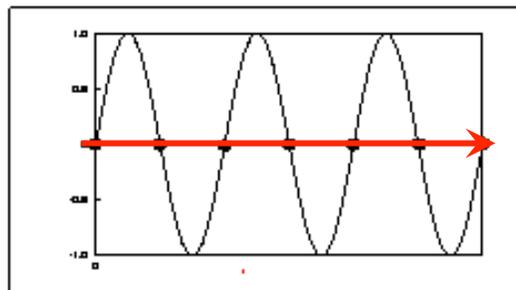
- SR = 20,000 Hz
- Nyquist Frequency = 10,000 Hz
- $f = 10,000$  Hz (no aliasing)



## Nyquist Theorem and Aliasing

### ■ Graphical Example 2:

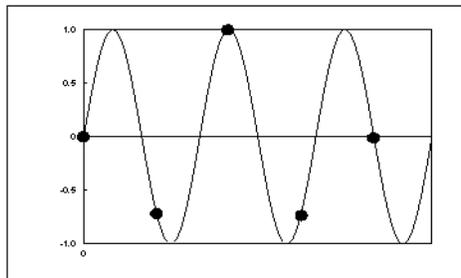
- BUT, if sample points fall on zero-crossings the sound is completely cancelled out



## Nyquist Theorem and Aliasing

### ■ Graphical Example 3:

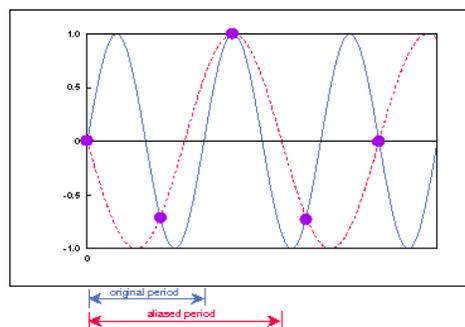
- SR = 20,000 Hz
- Nyquist Frequency = 10,000 Hz
- $f = 12,500$  Hz,  $f' = 7,500$



## Nyquist Theorem and Aliasing

### ■ Graphical Example 3:

- Fitting the simplest sine wave to the sampled points gives an aliased waveform (dotted line below):



## Method to reduce aliasing noise

