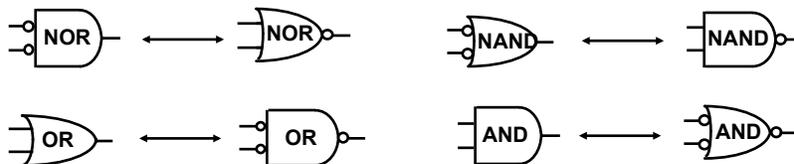


Canonical forms for Boolean logic

- Algebraic expressions to gates (lab 1)
- Canonical forms
- Incompletely specified functions
- Realizing two-level canonical forms

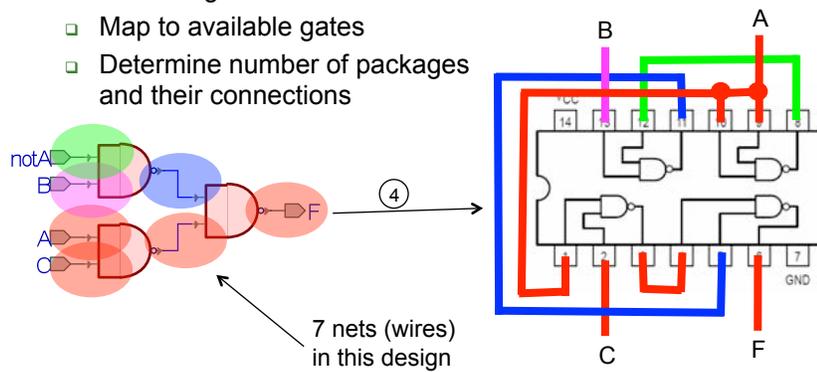
NAND, NOR, and de Morgan's theorem

- de Morgan's
 - Standard form: $A'B' = (A + B)'$ $A' + B' = (AB)'$
 - Inverted: $A + B = (A'B)'$ $(AB) = (A' + B)'$
 - AND with complemented inputs = NOR
 - OR with complemented inputs = NAND
 - OR = NAND with complemented inputs
 - AND = NOR with complemented inputs



Mapping truth tables to logic gates

- Given a truth table:
 - Write the Boolean expression
 - Minimize the Boolean expression
 - Draw as gates
 - Map to available gates
 - Determine number of packages and their connections

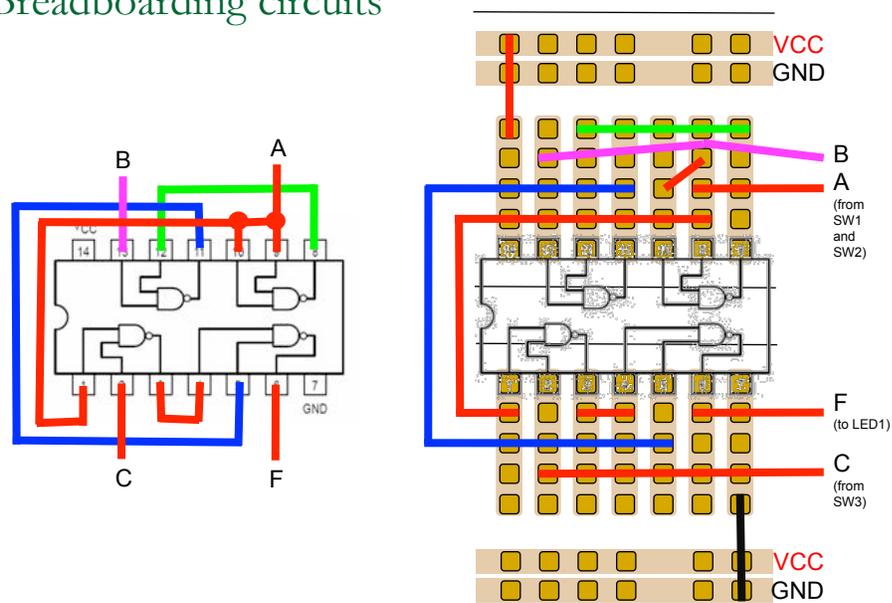


Spring 2010

CSE370 - IV - Canonical Forms

3

Breadboarding circuits

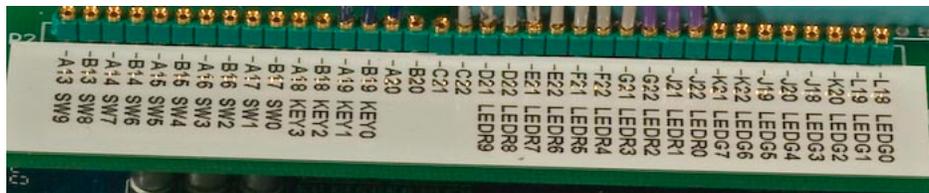
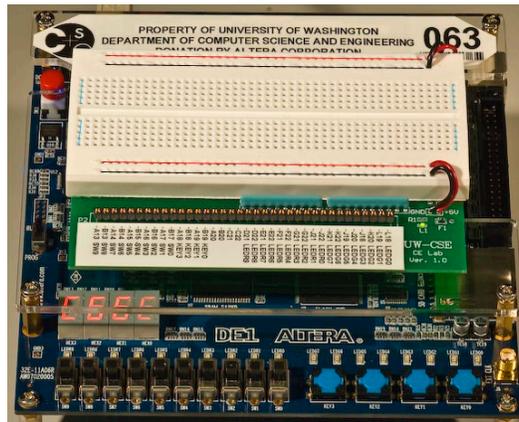


Spring 2010

CSE370 - IV - Canonical Forms

4

Lab 1 equipment



Random logic

- Too hard to figure out exactly what gates to use
 - map from logic to NAND/NOR networks
 - determine minimum number of packages
 - slight changes to logic function could decrease cost
- Changes too difficult to realize
 - need to rewire parts
 - may need new parts
 - design with spares (few extra inverters and gates on every board)
- Need higher levels of integration to keep costs down
 - cost directly related to number of devices and their pins

Regular logic

- Need to make design faster
- Need to make engineering changes easier to make
- Simpler for designers to understand and map to functionality
 - harder to think in terms of specific gates
 - easier to think in terms of larger multi-purpose blocks

Canonical forms

- Truth table is the unique signature of a Boolean function
- The same truth table can have many gate realizations
 - we've seen this already
 - depends on how good we are at Boolean simplification
- Canonical forms
 - standard forms for a Boolean expression
 - we all come up with the same expression

Sum-of-products canonical forms

- Also known as disjunctive normal form
- Also known as minterm expansion

				$F = 001 \quad 011 \quad 101 \quad 110 \quad 111$
				$F = A'B'C + A'BC + AB'C + ABC' + ABC$
A	B	C	F	F'
0	0	0	0	1
0	0	1	1	0
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	0

$F' = A'B'C' + A'BC' + AB'C'$

Sum-of-products canonical form (cont'd)

- Product term (or minterm)
 - ANDed product of literals – input combination for which output is true
 - each variable appears exactly once, true or inverted (but not both)

A	B	C	minterms	
0	0	0	A'B'C'	m0
0	0	1	A'B'C	m1
0	1	0	A'BC'	m2
0	1	1	A'BC	m3
1	0	0	AB'C'	m4
1	0	1	AB'C	m5
1	1	0	ABC'	m6
1	1	1	ABC	m7

short-hand notation for minterms of 3 variables

F in canonical form:

$$\begin{aligned}
 F(A, B, C) &= \Sigma m(1,3,5,6,7) \\
 &= m1 + m3 + m5 + m6 + m7 \\
 &= A'B'C + A'BC + AB'C + ABC' + ABC
 \end{aligned}$$

canonical form \neq minimal form

$$\begin{aligned}
 F(A, B, C) &= A'B'C + A'BC + AB'C + ABC + ABC' \\
 &= (A'B' + A'B + AB' + AB)C + ABC' \\
 &= ((A' + A)(B' + B))C + ABC' \\
 &= C + ABC' \\
 &= ABC' + C \\
 &= AB + C
 \end{aligned}$$

Product-of-sums canonical form

- Also known as conjunctive normal form
- Also known as maxterm expansion

				$F =$	000	010	100
				$F =$	$(A + B + C)$	$(A + B' + C)$	$(A' + B + C)$
A	B	C	F	F'			
0	0	0	0	1			
0	0	1	1	0			
0	1	0	0	1			
0	1	1	1	0			
1	0	0	0	1			
1	0	1	1	0			
1	1	0	1	0			
1	1	1	1	0			

$$F' = (A + B + C) (A + B' + C) (A' + B + C) (A' + B' + C) (A' + B' + C)$$

Product-of-sums canonical form (cont'd)

- Sum term (or maxterm)
 - ORed sum of literals – input combination for which output is false
 - each variable appears exactly once, true or inverted (but not both)

A	B	C	maxterms	
0	0	0	A+B+C	M0
0	0	1	A+B+C'	M1
0	1	0	A+B'+C	M2
0	1	1	A+B'+C'	M3
1	0	0	A'+B+C	M4
1	0	1	A'+B+C'	M5
1	1	0	A'+B'+C	M6
1	1	1	A'+B'+C'	M7

F in canonical form:

$$\begin{aligned} F(A, B, C) &= \Pi M(0,2,4) \\ &= M0 \cdot M2 \cdot M4 \\ &= (A + B + C) (A + B' + C) (A' + B + C) \end{aligned}$$

canonical form \neq minimal form

$$\begin{aligned} F(A, B, C) &= (A + B + C) (A + B' + C) (A' + B + C) \\ &= (A + B + C) (A + B' + C) \\ &\quad (A + B + C) (A' + B + C) \\ &= (A + C) (B + C) \end{aligned}$$

short-hand notation for maxterms of 3 variables

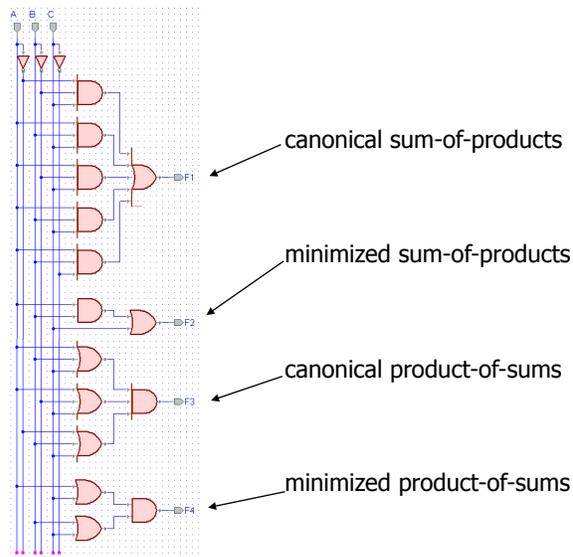
S-o-P, P-o-S, and de Morgan's theorem

- Complement of function in sum-of-products form
 - $F' = A'B'C' + A'BC' + AB'C'$
- Complement again and apply de Morgan's and get the product-of-sums form
 - $(F')' = (A'B'C' + A'BC' + AB'C')'$
 - $F = (A + B + C)(A + B' + C)(A' + B + C)$
- Complement of function in product-of-sums form
 - $F' = (A + B + C')(A + B' + C')(A' + B + C')(A' + B' + C)(A' + B' + C')$
- Complement again and apply de Morgan's and get the sum-of-product form
 - $(F')' = ((A + B + C')(A + B' + C')(A' + B + C')(A' + B' + C)(A' + B' + C'))'$
 - $F = A'B'C' + A'BC' + AB'C' + ABC' + ABC$

Mapping between canonical forms

- Minterm to maxterm conversion
 - use maxterms whose indices do not appear in minterm expansion
 - e.g., $F(A,B,C) = \Sigma m(1,3,5,6,7) = \Pi M(0,2,4)$
- Maxterm to minterm conversion
 - use minterms whose indices do not appear in maxterm expansion
 - e.g., $F(A,B,C) = \Pi M(0,2,4) = \Sigma m(1,3,5,6,7)$
- Minterm expansion of F to minterm expansion of F'
 - use minterms whose indices do not appear
 - e.g., $F(A,B,C) = \Sigma m(1,3,5,6,7)$ $F'(A,B,C) = \Sigma m(0,2,4)$
- Maxterm expansion of F to maxterm expansion of F'
 - use maxterms whose indices do not appear
 - e.g., $F(A,B,C) = \Pi M(0,2,4)$ $F'(A,B,C) = \Pi M(1,3,5,6,7)$

Four alternative two-level implementations of $F = AB + C$



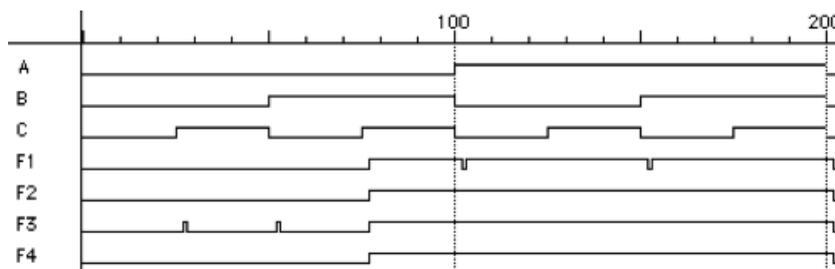
Spring 2010

CSE370 - IV - Canonical Forms

15

Waveforms for the four alternatives

- Waveforms are essentially identical
 - except for timing hazards (glitches)
 - delays almost identical (modeled as a delay per level, not type of gate or number of inputs to gate)

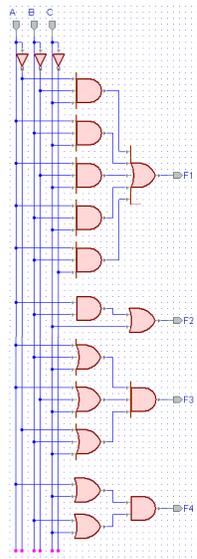


Spring 2010

CSE370 - IV - Canonical Forms

16

Four alternative two-level implementations of $F = AB + C$



	Transistors (NOT = 2)	Delay (approx) (NOT = 1)	Hazards
F1	5 3-input NANDs 1 5-input NAND $5*6 + 1*10 = 40$	2 levels $3^2 + 5^2 = 34$	yes
F2	2 2-input NANDs $2*4 = 8$	2 levels $2^2 + 2^2 = 8$	no
F3	4 3-input NANDs $4*6 = 24$	2 levels $3^2 + 3^2 = 18$	yes
F4	3 2-input NANDs $3*4 = 12$	2 levels $2^2 + 2^2 = 8$	no

Spring 2010

CSE370 - IV - Canonical Forms

17

Incompletely specified functions

- Example: binary coded decimal increment by 1

- BCD digits encode the decimal digits 0 – 9 in the bit patterns 0000 – 1001

A	B	C	D	W	X	Y	Z
0	0	0	0	0	0	0	1
0	0	0	1	0	0	1	0
0	0	1	0	0	0	1	1
0	0	1	1	0	1	0	0
0	1	0	0	0	1	0	1
0	1	0	1	0	1	1	0
0	1	1	0	0	1	1	1
0	1	1	1	1	0	0	0
1	0	0	0	1	0	0	1
1	0	0	1	0	0	0	0
1	0	1	0	X	X	X	X
1	0	1	1	X	X	X	X
1	1	0	0	X	X	X	X
1	1	0	1	X	X	X	X
1	1	1	0	X	X	X	X
1	1	1	1	X	X	X	X

Annotations:

- off-set of W (points to W=0 for 0001, 0010, 0011, 0100, 0101, 0110, 0111)
- on-set of W (points to W=1 for 1000, 1001)
- don't care (DC) set of W (points to W=X for 1010-1111)

these inputs patterns should never be encountered in practice – "don't care" about output values in these cases – might be useful in minimization

Spring 2010

CSE370 - IV - Canonical Forms

18

Notation for incompletely specified functions

- Don't cares and canonical forms
 - so far, only represented on-set
 - also represent don't-care-set
 - need two of the three sets (on-set, off-set, dc-set)

- Canonical representations of the BCD increment by 1 function:
 - $Z = m_0 + m_2 + m_4 + m_6 + m_8 + d_{10} + d_{11} + d_{12} + d_{13} + d_{14} + d_{15}$
 - $Z = \Sigma [m(0,2,4,6,8) + d(10,11,12,13,14,15)]$

 - $Z = M_1 \cdot M_3 \cdot M_5 \cdot M_7 \cdot M_9 \cdot D_{10} \cdot D_{11} \cdot D_{12} \cdot D_{13} \cdot D_{14} \cdot D_{15}$
 - $Z = \Pi [M(1,3,5,7,9) \cdot D(10,11,12,13,14,15)]$