

Intro to Digital Design

FSM Design, MUXes, Adders

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Relevant Course Information

- ❖ Lab 6 – Connecting multiple FSMs in Tug of War game
 - *Bigger* step up in difficulty from Lab 5
 - Putting together complex system – interconnections!
 - Bonus points for smaller resource usage

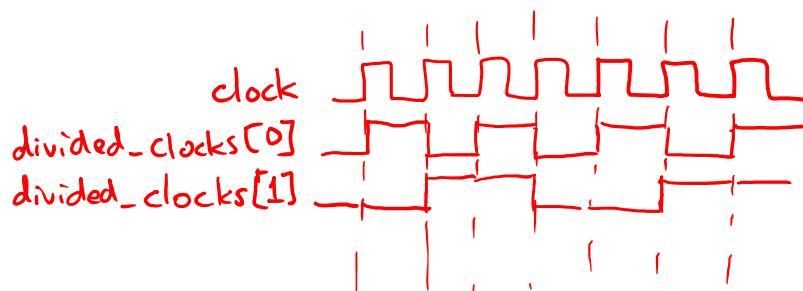
Clock Divider (not for simulation)

CLOCK_50: 50 MHz clock on your FPGA

- ❖ Why/how does this work?

```
// divided_clocks[0]=25MHz, [1]=12.5Mhz, ...
module clock_divider (clock, divided_clocks);
    input logic          clock;
    output logic [31:0] divided_clocks;
    initial
        divided_clocks = 0;

    always_ff @ (posedge clock)
        divided_clocks <= divided_clocks + 1;
endmodule
```



$$\begin{aligned} \text{divided_clocks} = & 32\text{'b}0\dots000 \\ & 32\text{'b}0\dots001 \\ & 0\dots010 \\ & 0\dots011 \\ & 0\dots100 \end{aligned}$$

→ use divided_clocks [#]
everywhere else in your
system

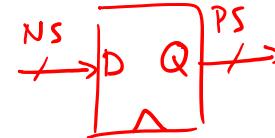
changes every fourth trigger
changes every other trigger
changes every clock trigger

Outline

- ❖ **FSM Design**
- ❖ Multiplexors
- ❖ Adders

FSM Design Process

- 1) Understand the problem 
- 2) Draw the state diagram
311 knowledge
- 3) Use state diagram to produce state table
read off transitions
- 4) Implement the combinational control logic
CL + SL
gates + registers



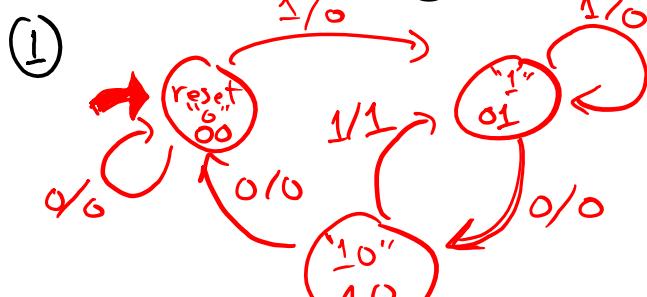
Practice: String Recognizer FSM

- ① Draw the FSM
- ② Truth Table
- ③ Simplify Logic
- ④ Circuit Diagram

- ❖ Recognize the string 101 with the following behavior

- Input: 1 0 0 1 0 1 0 1 1 0 0 1 0
- Output: 0 0 0 0 0 1 0 1 0 0 0 0 0

- ❖ State diagram to implementation:



②

PS	In	NS	Out
00	0	00	0
00	1	01	0
01	0	10	0
01	1	01	0
10	0	00	1
10	1	01	1
11	0	X	X
11	1	X	X

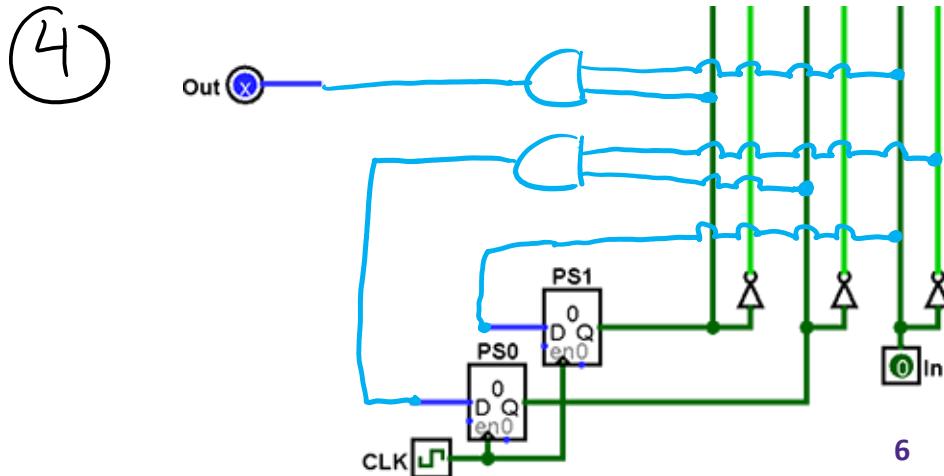
③

PS	In	NS	Out
00	0	01	0
00	1	00	0
01	0	00	0
01	1	10	0
10	0	00	0
10	1	01	0

PS	In	NS	Out
00	0	00	0
00	1	00	0
01	0	00	0
01	1	11	1
11	0	11	1
11	1	00	0

PS	In	NS	Out
00	0	00	0
00	1	00	0
01	0	00	0
01	1	00	0
11	0	X	0
11	1	X	1

$NS_1 = \overline{In} \cdot PS_0$
 $NS_0 = In$
 $Out = In \cdot PS_1$



HDL Organization

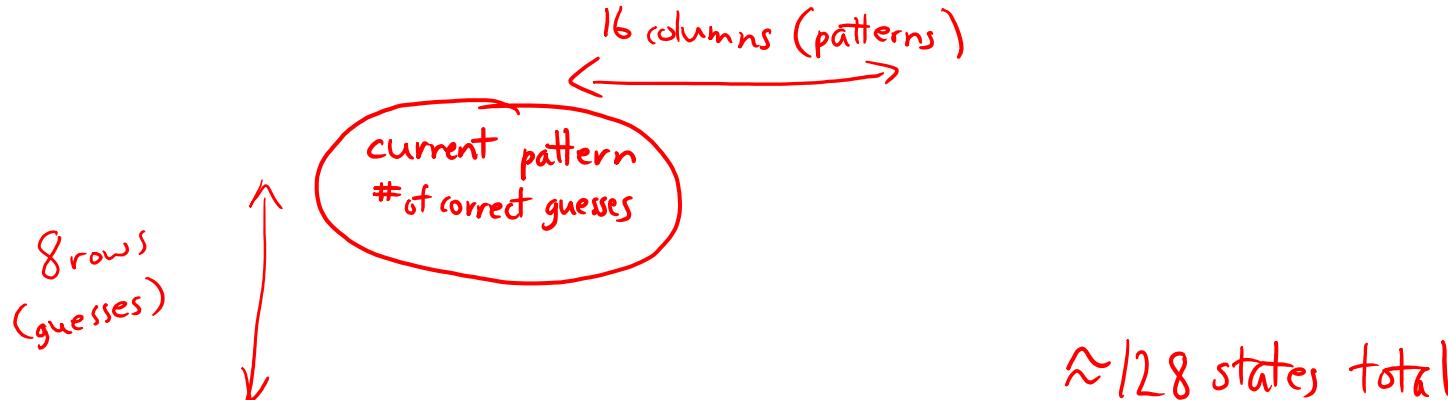
- ❖ Most problems are best solved with multiple pieces – how to best organize your system and code?
- ❖ Everything is computed in parallel
 - We use routing elements (next lecture) to select between (or ignore) multiple outcomes/parts
 - This is why we use block diagrams and waveforms
- ❖ A module is not a *function*, it is closest to a *class*
 - Something that you *instantiate*, not something that you *call* – hardware cannot appear and disappear spontaneously
 - Should treat modules as *resource managers* rather than temporary helpers
 - This can include having internal modules

Subdividing FSMs Example

❖ “Psychic Tester”

- Machine generates a 4-bit pattern $2^4 = 16$ patterns
- User tries to guess 8 patterns in a row to be deemed psychic
0-7 correct guesses so far (8 total)

❖ States?



Example: Plan First with Block Diagram

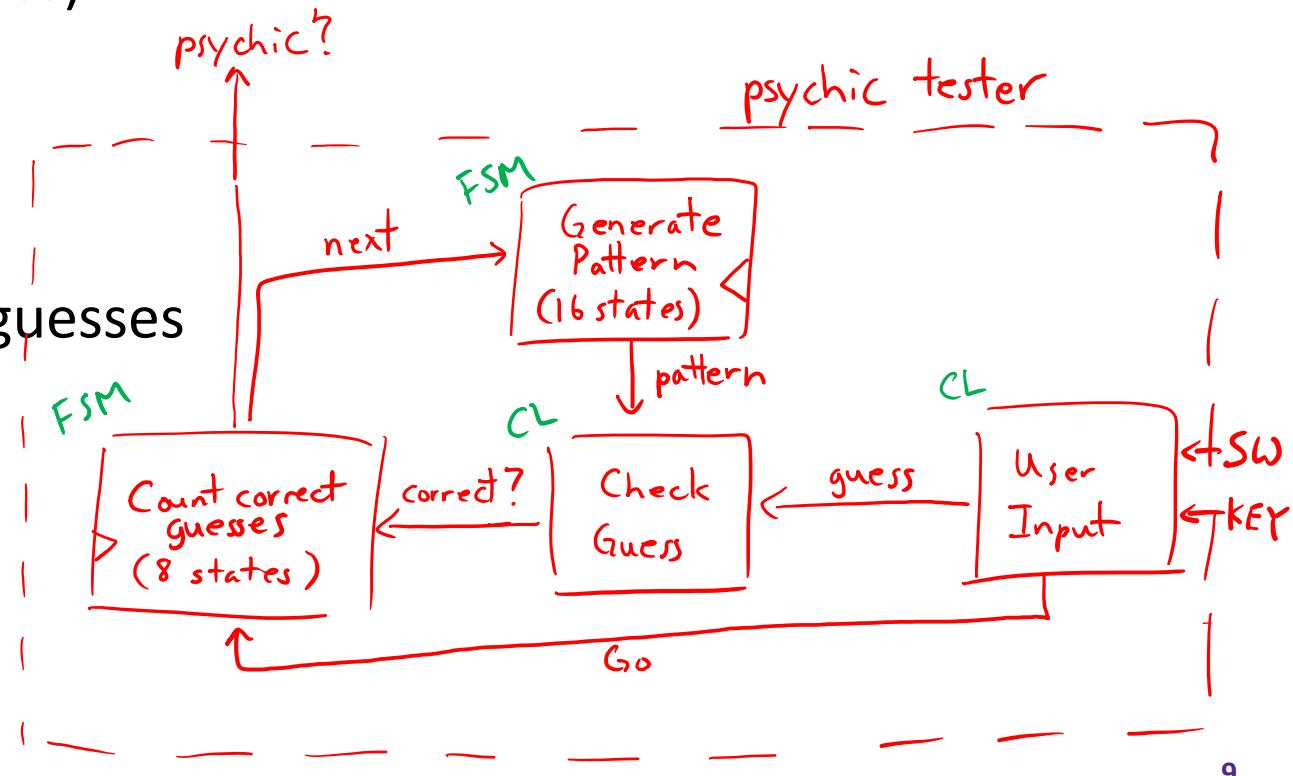
❖ Pieces?

- Generate/pick pattern

- User input (guess)

- Check guess

- Count correct guesses



Example: Blocks → Modules

❖ Pieces?

- Generate/pick pattern
 - module genPatt(pattern, next, clock);

- User input (guess)
 - module userIn(guess, enable, KEY);

- Check guess
 - module checkGuess(correct, guess, pattern);

- Count correct guesses
 - module countRight(psychic, next, correct, enable, clock);

Example: Implementation & Testing

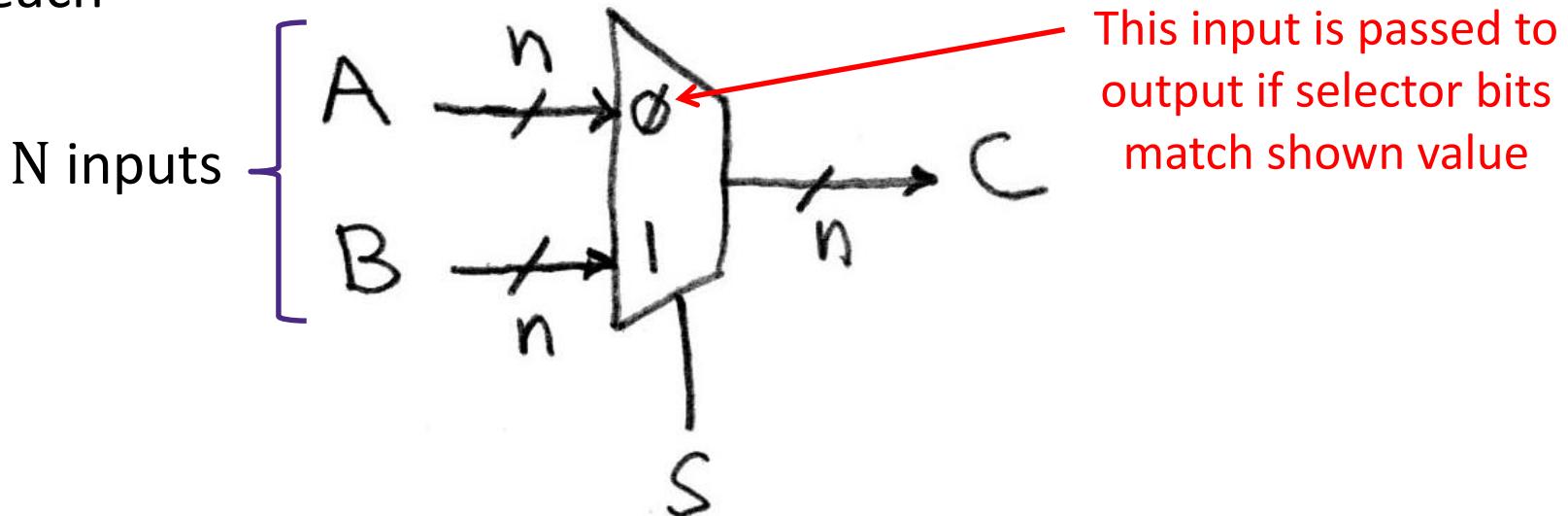
- 1) Create individual submodules
- 2) Create submodules test benches – test as usual
 - CL – run through all input combinations
 - SL – take every transition that you care about
- 3) Create top-level module
 - Create instance of each submodule
 - Create wires/nets to connect signals between submodules, inputs, and outputs
- 4) Create top-level test bench
 - Goal is to check the interconnections between submodules – does input/state change in one submodule trigger the expected change in other submodules?

Outline

- ❖ FSM Design
- ❖ Multiplexors
- ❖ Adders

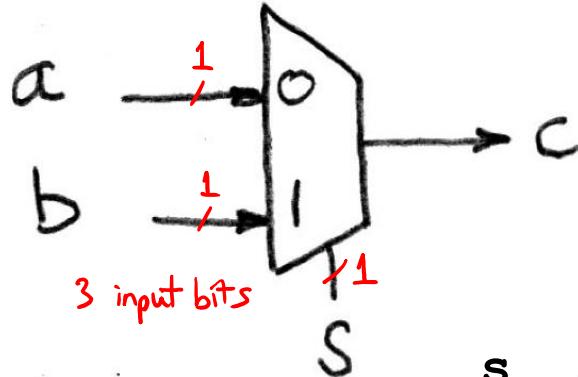
Data Multiplexor

- ❖ Multiplexor (“MUX”) is a *selector*
 - Direct one of many ($N = 2^s$) n -bit wide inputs onto output
 - Called a n -bit, N -to-1 MUX
 - \uparrow bus widths
 - \uparrow possible selections
- ❖ Example: n -bit 2-to-1 MUX
 - Input S (s bits wide) selects between two inputs of n bits each



Review: Implementing a 1-bit 2-to-1 MUX

❖ Schematic:



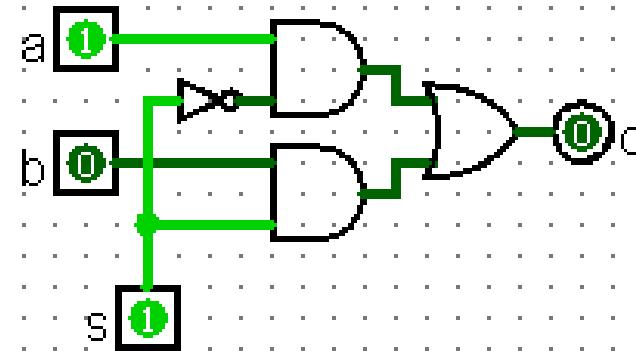
❖ Truth Table:

s	a	b	c
0	0	0	0
0	0	1	0
0	1	0	1
0	1	1	1
1	0	0	0
1	0	1	1
1	1	0	0
1	1	1	1

❖ Boolean Algebra:

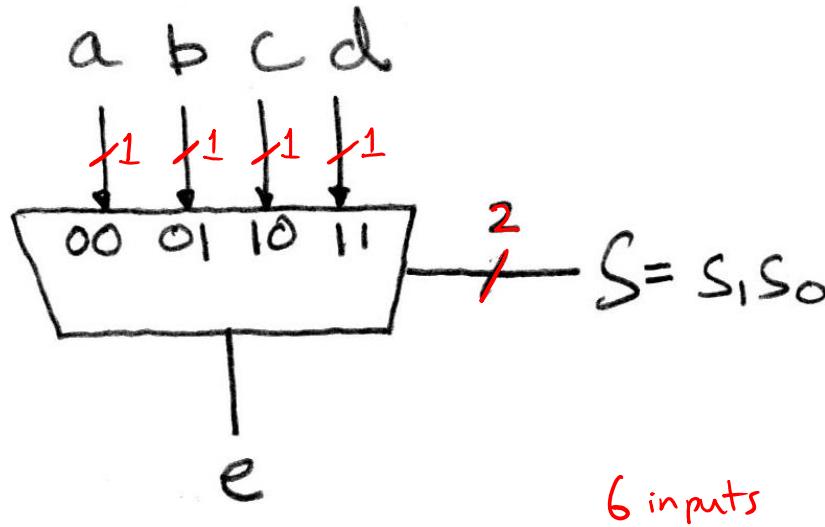
$$\begin{aligned}c &= \bar{s}a\bar{b} + \bar{s}ab + s\bar{a}\bar{b} + sab \\&= \bar{s}(a\bar{b} + ab) + s(\bar{a}\bar{b} + ab) \\&= \bar{s}(a(\bar{b} + b)) + s((\bar{a} + a)b) \\&= \bar{s}(a(1) + s((1)b) \\&= \bar{s}a + sb\end{aligned}$$

❖ Circuit Diagram:



1-bit 4-to-1 MUX

- ❖ Schematic:



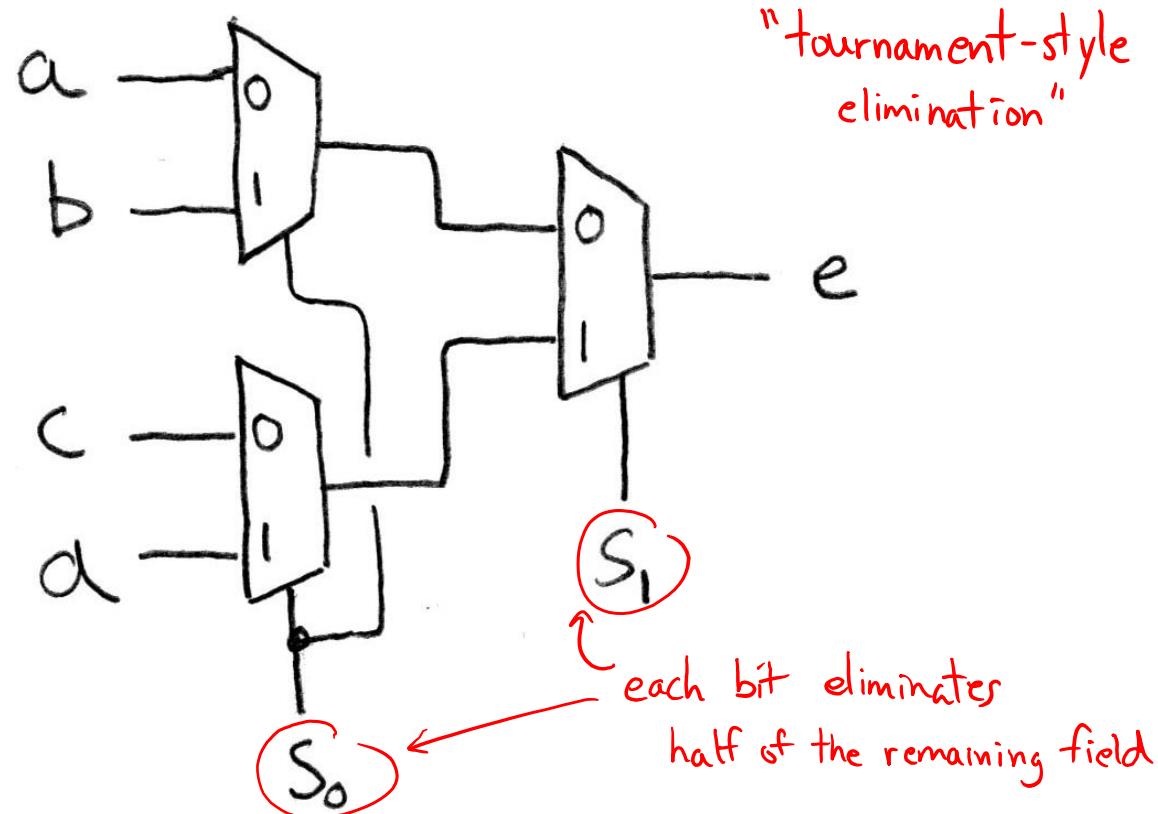
- ❖ Truth Table: How many rows? 2^6

- ❖ Boolean Expression:

$$e = \bar{s}_1 \bar{s}_0 a + \bar{s}_1 s_0 b + s_1 \bar{s}_0 c + s_1 s_0 d$$

1-bit 4-to-1 MUX

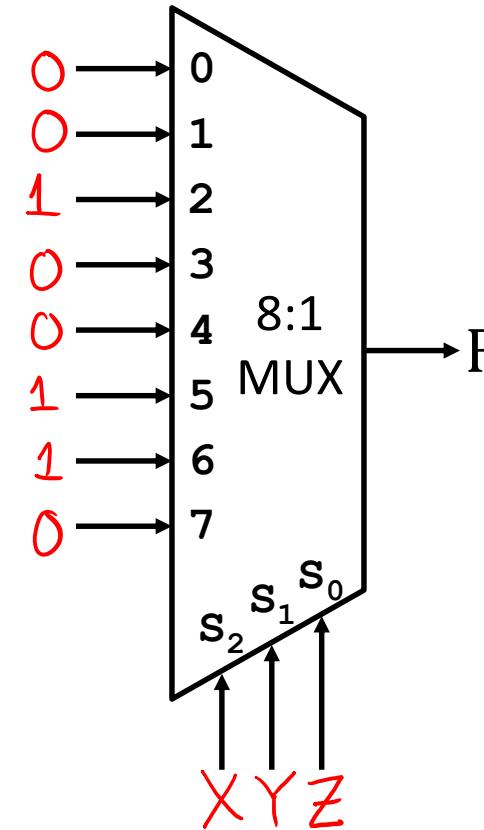
- ❖ Can we leverage what we've previously built?
 - Alternative hierarchical approach:



Multiplexers in General Logic

- ❖ Implement $F = X\bar{Y}Z + Y\bar{Z}$ with a 8:1 MUX

$X \bar{Y} Z$	F
0 0 0	0
0 0 1	0
0 1 0	1
0 1 1	0
1 0 0	0
1 0 1	1
1 1 0	1
1 1 1	0



Technology Break

Outline

- ❖ FSM Design
- ❖ Multiplexors
- ❖ Adders

Review: Unsigned Integers

- ❖ Unsigned values follow the standard base 2 system
 - $b_7b_6b_5b_4b_3b_2b_1b_0 = b_72^7 + b_62^6 + \dots + b_12^1 + b_02^0$
- ❖ In n bits, represent integers 0 to $2^n - 1$
- ❖ Add and subtract using the normal “carry” and “borrow” rules, just in binary

	Carry 1 1 1			borrow 1 1 1 1
63	00111111	64	01000000	
+ 8	+ 00001000	- 8	- 00001000	
71	01000111	56	00111000	

Review: Two's Complement (Signed)

b_{w-1} has weight -2^{w-1} , other bits have usual weights $+2^i$



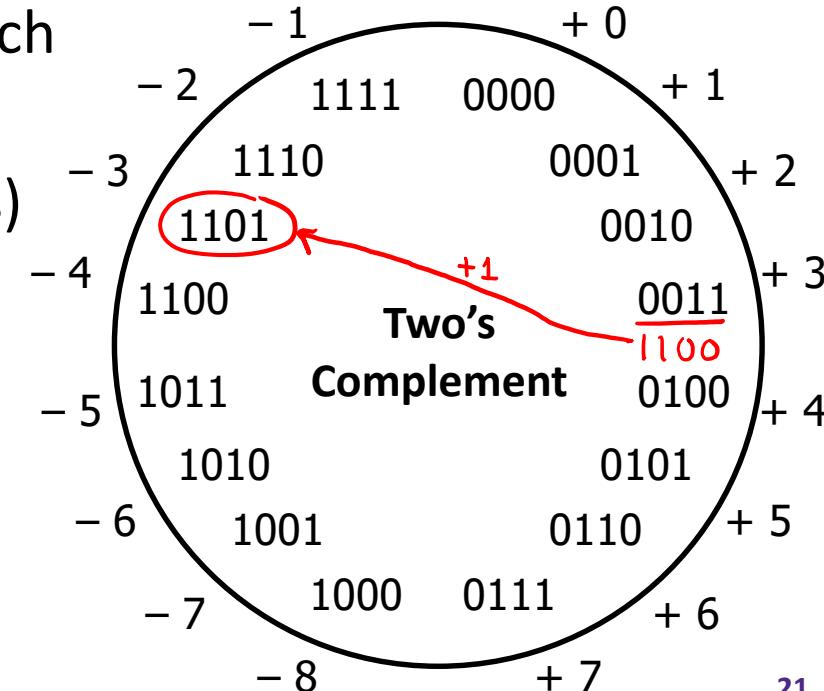
Properties:

- In n bits, represent integers -2^{n-1} to $2^{n-1} - 1$
- Positive number encodings match unsigned numbers
- Single zero (encoding = all zeros)

Negation procedure:

- Take the bitwise complement and then add one

$$(\sim x + 1 == -x)$$



Addition and Subtraction in Hardware

- ❖ The same bit manipulations work for both unsigned and two's complement numbers!
 - Perform subtraction via adding the negated 2nd operand:
$$A - B = A + (-B) = A + (\sim B) + 1$$
- ❖ 4-bit examples:

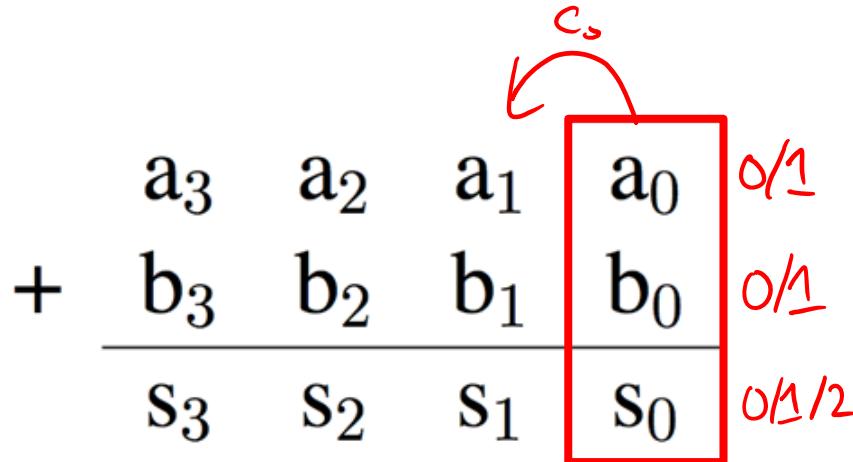
	Two's	Un
0 0 1 0	+2	2
+ 1 1 0 0	-4	12
<hr/>	-2	14

0 1 1 0	+6	6
- 0 0 1 0	+2	2
<hr/>	+ 1 1 0 1	
<hr/>	0 1 0 0	4

	Two's	Un
1 0 0 0	-8	8
+ 0 1 0 0	+4	4
<hr/>	-4	12

1 1 1 1	-1	15
- 1 1 1 0	-2	14
<hr/>	+ 0 0 0 1	
<hr/>	0 0 0 1	1

Half Adder (1 bit)

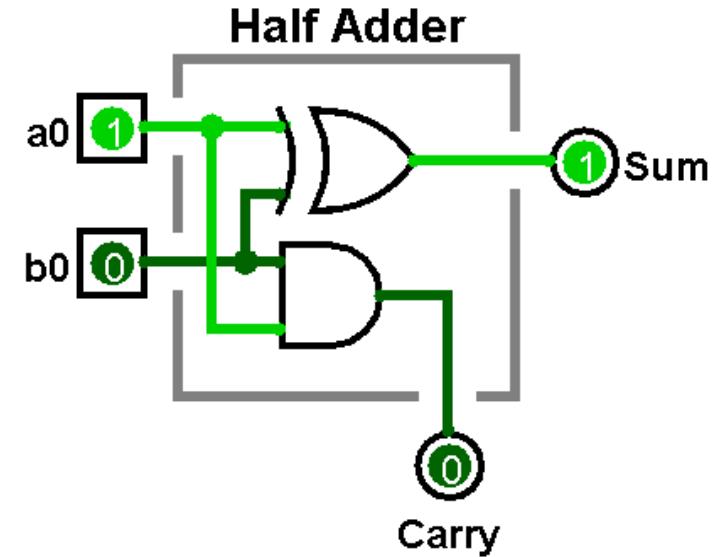


$$\text{Carry} = a_0 b_0$$

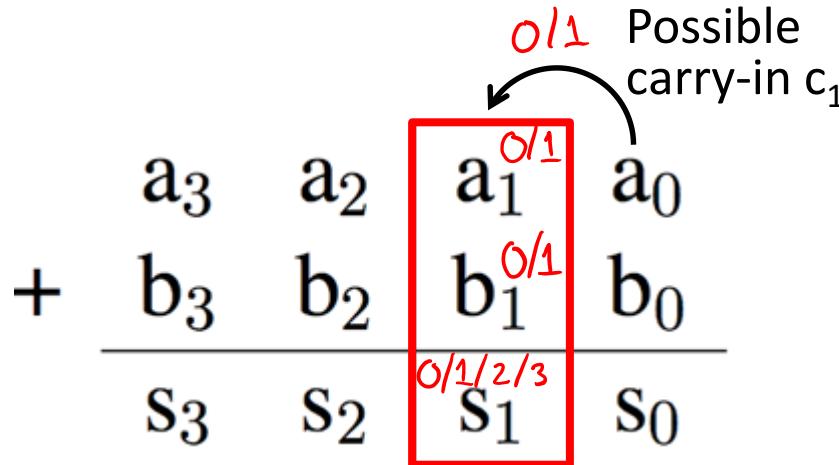
$$\text{Sum} = a_0 \oplus b_0$$

a_0	b_0	c_1	s_0
0	0	0	0
0	1	0	1
1	0	0	1
1	1	1	0

Carry-out bit



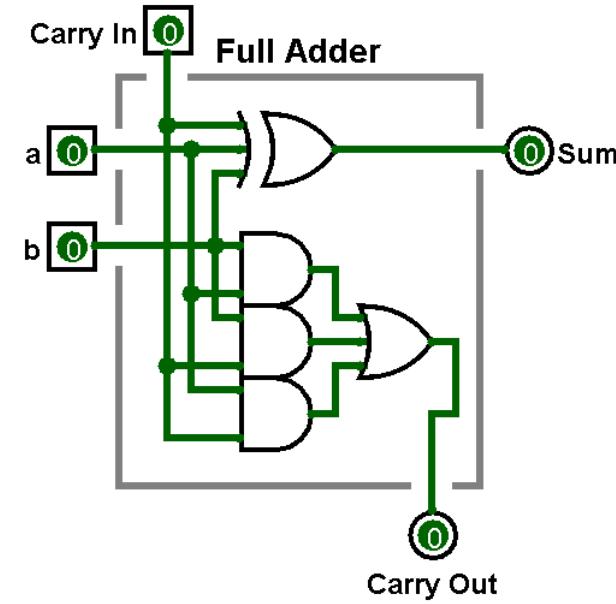
Full Adder (1 bit)



Carry-in Carry-out

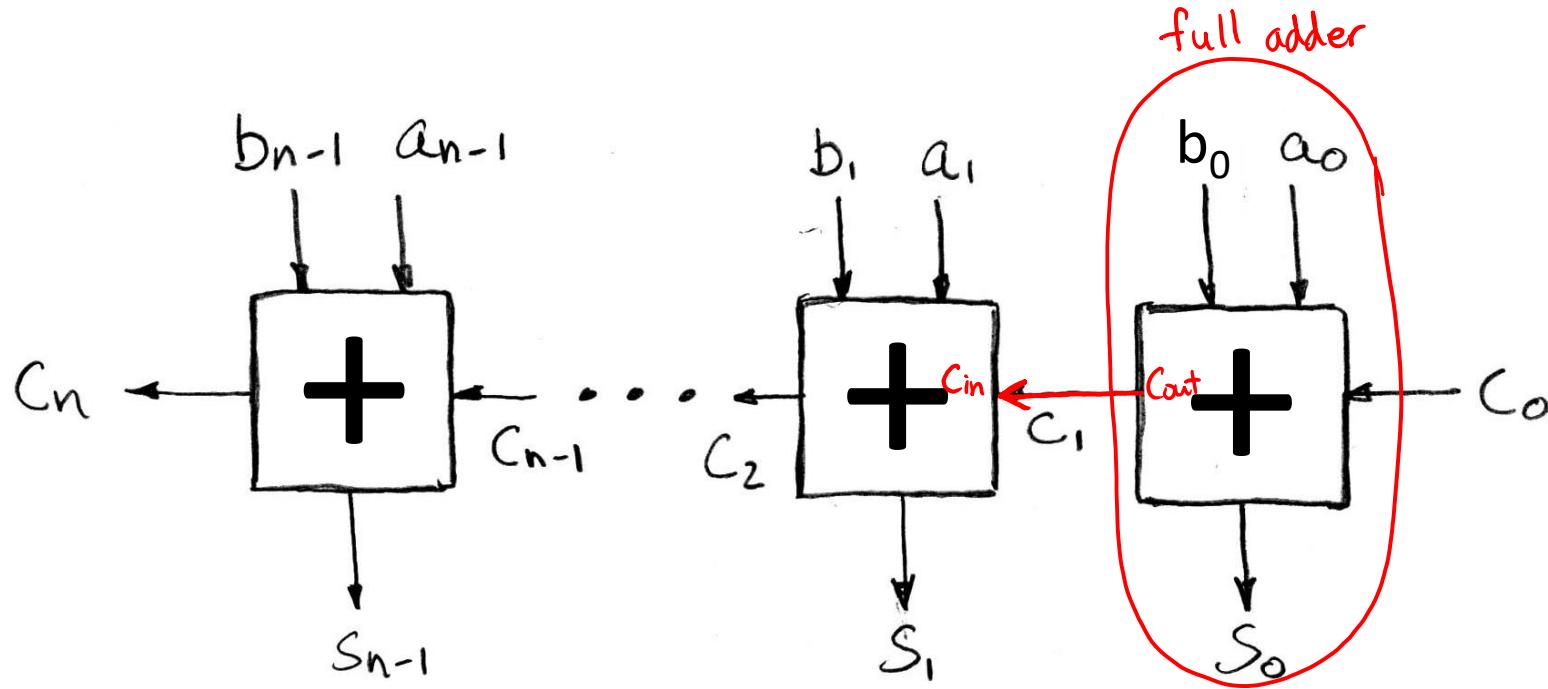
c_i	a_i	b_i	c_{i+1}	s_i
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

$$\begin{aligned}
 s_i &= \text{XOR}(a_i, b_i, c_i) \\
 c_{i+1} &= \text{MAJ}(a_i, b_i, c_i) \\
 &= a_i b_i + a_i c_i + b_i c_i
 \end{aligned}$$



Multi-Bit Adder (N bits)

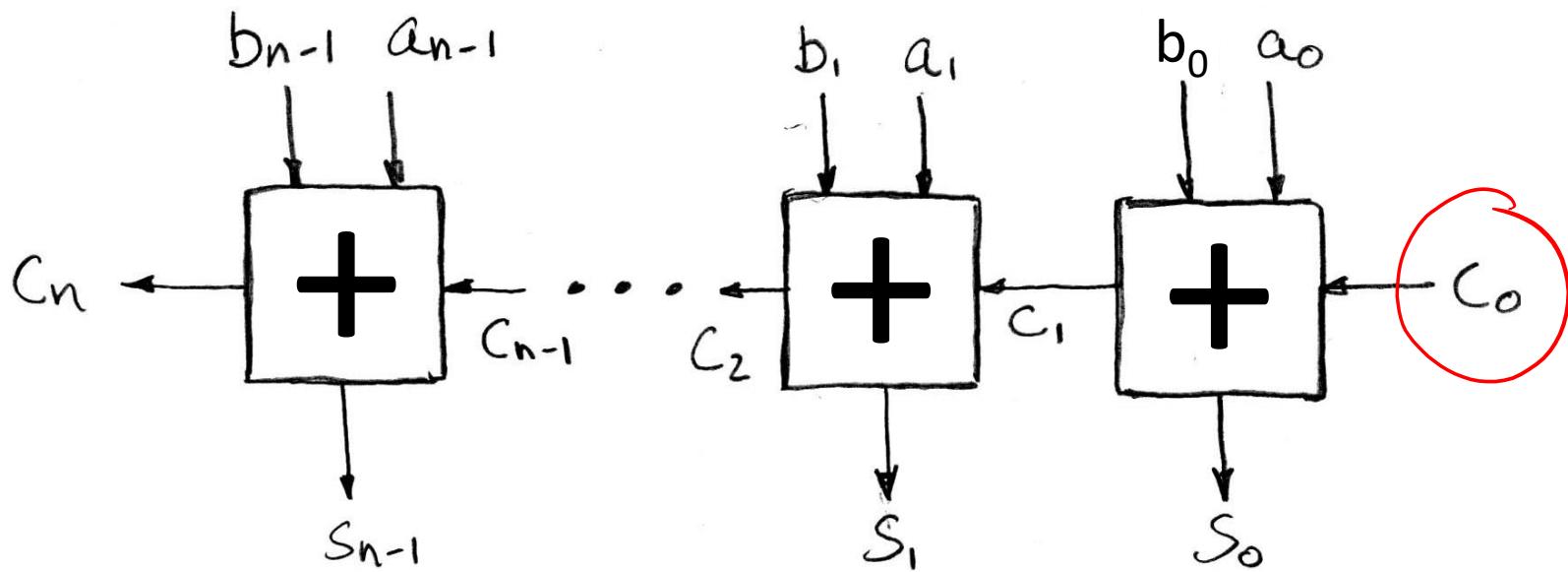
- ❖ Chain 1-bit adders by connecting CarryOut_i to CarryIn_{i+1} :



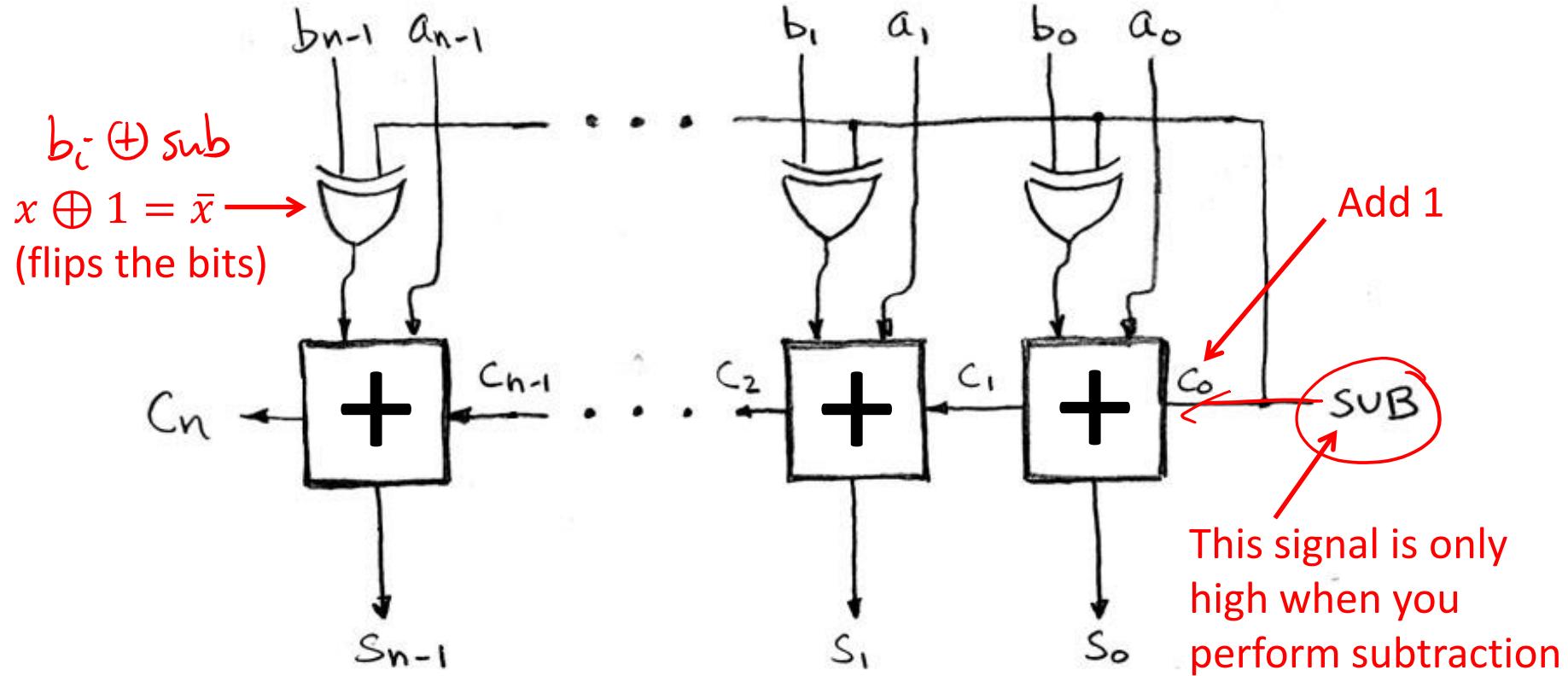
Subtraction?

- ❖ Can we use our multi-bit adder to do subtraction?
 - Flip the bits and add 1?
 - $X \oplus 1 = \bar{X}$
 - CarryIn₀ (using full adder in all positions)

$$\begin{array}{l} x \& 0 = 0 \\ x \& 1 = x \\ x \mid 0 = x \\ x \mid 1 = 1 \\ \hline x \wedge 0 = x \\ x \wedge 1 = \bar{x} \end{array}$$



Multi-bit Adder/Subtractor



Detecting Arithmetic Overflow

- ❖ **Overflow:** When a calculation produces a result that can't be represented in the current encoding scheme
 - Integer range limited by fixed width
 - Can occur in both the positive and negative directions

❖ Unsigned Overflow

- Result of add/sub is > UMax or < Umin
 $0b11\dots1$ $0b00\dots0$

❖ Signed Overflow

- Result of add/sub is > TMax or < TMin
 $0b01\dots1$ $0b10\dots0$
- $(+) + (+) = (-)$ or $(-) + (-) = (+)$

Signed Overflow Examples

Two's

$$\begin{array}{r} 0 \ 1 \ 0 \ 1 \\ + 0 \ 0 \ 1 \ 1 \\ \hline 1 \ 0 \ 0 \ 0 \end{array}$$

-8 overflow

Two's

$$\begin{array}{r} 0 \ 1 \ 0 \ 1 \\ + 0 \ 0 \ 1 \ 0 \\ \hline 0 \ 1 \ 1 \ 1 \end{array}$$

7

Two's

$$\begin{array}{r} 1 \ 0 \ 0 \ 1 \\ + 1 \ 1 \ 1 \ 0 \\ \hline 0 \ 1 \ 1 \ 1 \end{array}$$

+7 overflow

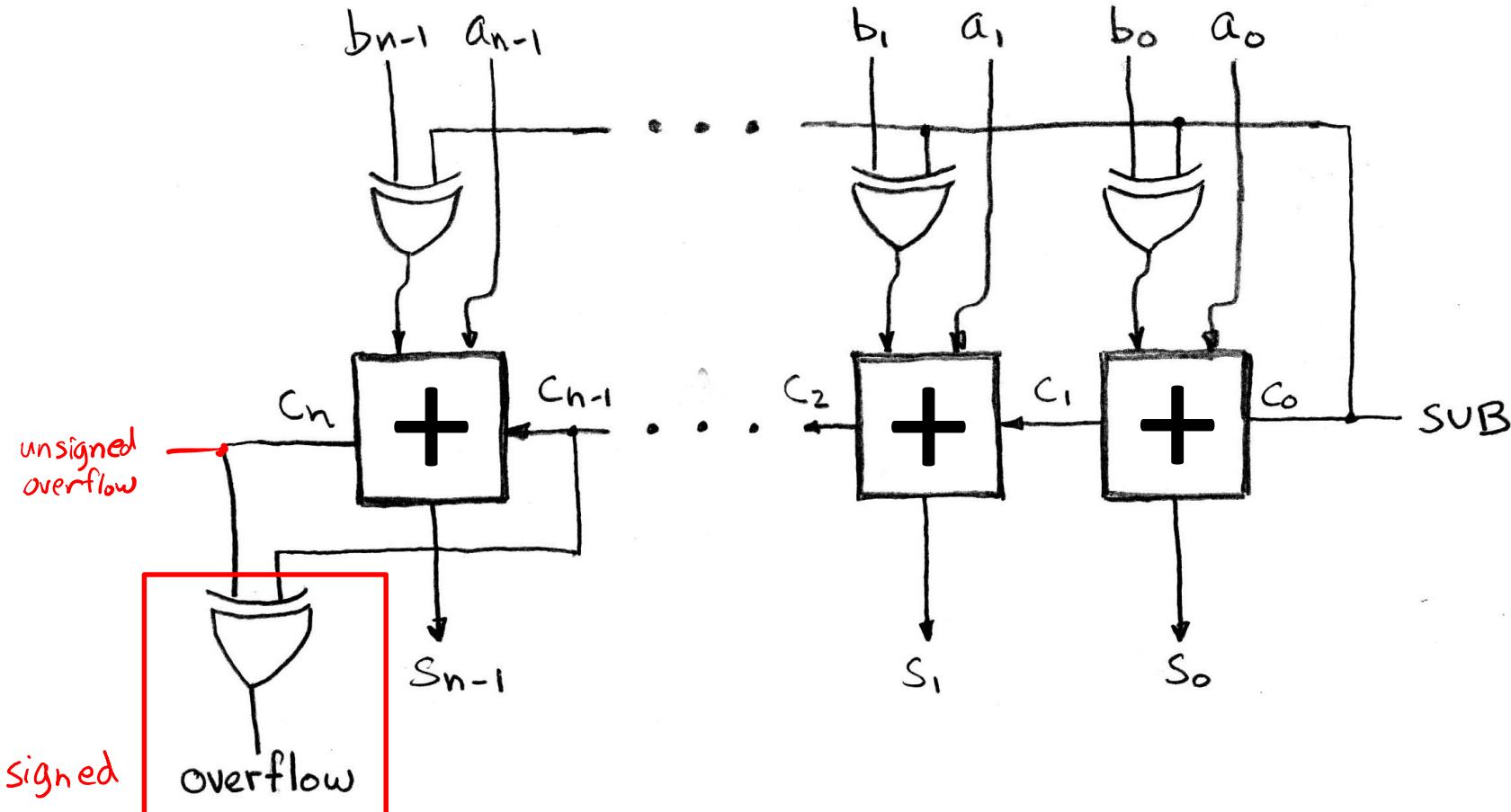
Two's

$$\begin{array}{r} 1 \ 1 \ 0 \ 0 \\ + 0 \ 1 \ 0 \ 0 \\ \hline 0 \ 0 \ 0 \ 0 \end{array}$$

0

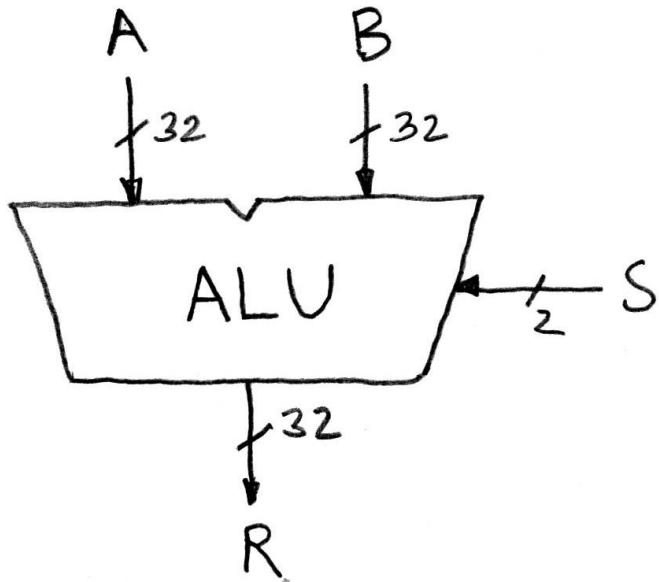
$$\boxed{\text{Overflow} = C_n \wedge C_{n-1}}$$

Multi-bit Adder/Subtractor with Overflow



Arithmetic and Logic Unit (ALU)

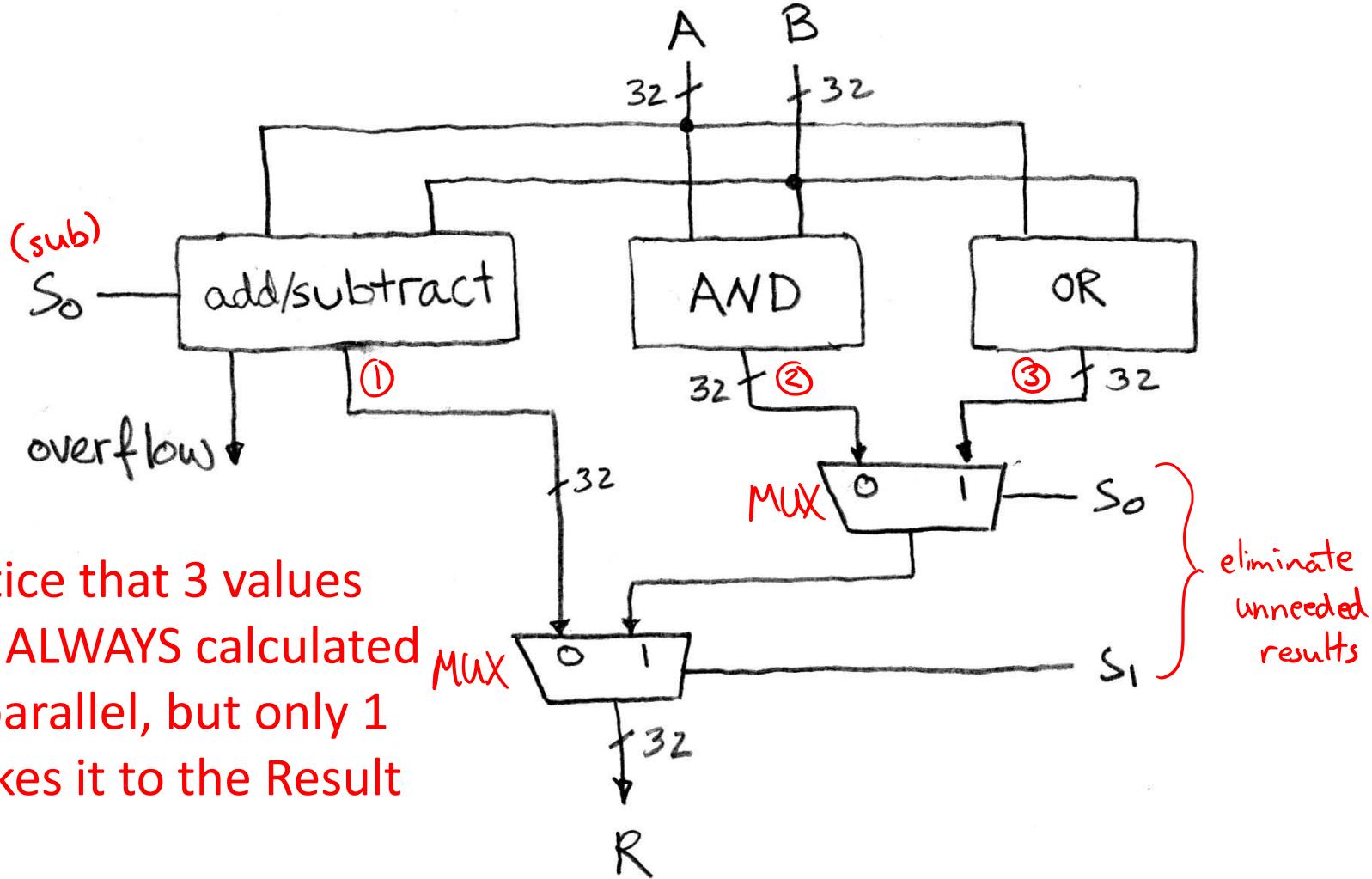
- ❖ Processors contain a special logic block called the “Arithmetic and Logic Unit” (ALU)
 - Here’s an easy one that does ADD, SUB, bitwise AND, and bitwise OR (for 32-bit numbers)
- ❖ Schematic:



"arbitrary but affects implementation"

when $S=00$, $R = A+B$
when $S=01$, $R = A-B$
when $S=10$, $R = A \& B$
when $S=11$, $R = A | B$

Simple ALU Schematic



1-bit Adders in Verilog

- ❖ What's wrong with this?
 - Truncation!

```
module halfadd1 (s, a, b);  
    output logic s;  
    input logic a, b;  
    single bit  
    always_comb begin  
        s = a + b;  
    end  
endmodule
```

- ❖ Fixed:
 - Use of {sig, ..., sig} for *concatenation*

```
module halfadd2 (c, s, a, b);  
    output logic c, s;  
    input logic a, b;  
  
    always_comb begin  
        {c, s} = a + b;  
    end  
endmodule
```

could have been:
 $s = a \oplus b;$
 $c = a \& b;$

Order matters!

Ripple-Carry Adder in Verilog

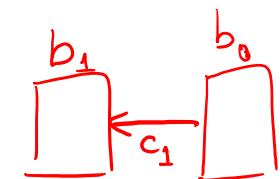
```
module fulladd (cout, s, cin, a, b);
    output logic cout, s;
    input  logic cin, a, b;

    always_comb begin
        {cout, s} = cin + a + b;
    end
endmodule
```

- ❖ Chain full adders?

```
module add2 (cout, s, cin, a, b);
    output logic cout; output logic [1:0] s;
    input  logic cin;  input  logic [1:0] a, b;
    logic c1;

    fulladd b1 (cout, s[1], c1, a[1], b[1]);
    fulladd b0 (c1,   s[0], cin, a[0], b[0]);
endmodule
```



Add/Sub in Verilog (parameterized)

❖ Variable-width add/sub (with overflow, carry)

```
module addN #(parameter N=32) (OF, CF, S, sub, A, B);
    output logic OF, CF; // overflow and carry "flags"
    output logic [N-1:0] S;
    input logic [N-1:0] A, B;
    logic [N-1:0] D; // possibly flipped B
    logic C2; // second-to-last carry-out

    always_comb begin
        D = B ^ {N{sub}}; // replication operator
        {C2, S[N-2:0]} = A[N-2:0] + D[N-2:0] + sub;
        {CF, S[N-1]} = A[N-1] + D[N-1] + C2;
        OF = CF ^ C2;
    end
endmodule
```

default value

sub; parameter changes bus widths

flip all bits of B if sub==1

- Here using OF = overflow flag, CF = carry flag
 - From condition flags in x86-64 processors

Add/Sub in Verilog (parameterized)

```
module addN_testbench ();
    parameter N = 4;
    logic sub;
    logic [N-1:0] A, B;
    logic OF, CF;
    logic [N-1:0] S;

    addN #(.(N)) dut (.OF, .CF, .S, .sub, .A, .B);

    initial begin
        #100; sub = 0; A = 4'b0101; B = 4'b0010; // 5 + 2
        #100; sub = 0; A = 4'b1101; B = 4'b1011; // -3 + -5
        #100; sub = 0; A = 4'b0101; B = 4'b0011; // 5 + 3
        #100; sub = 0; A = 4'b1001; B = 4'b1110; // -7 + -2
        #100; sub = 1; A = 4'b0101; B = 4'b1110; // 5 - (-2)
        #100; sub = 1; A = 4'b1101; B = 4'b0101; // -3 - 5
        #100; sub = 1; A = 4'b0101; B = 4'b1101; // 5 - (-3)
        #100; sub = 1; A = 4'b1001; B = 4'b0010; // -7 - 2
        #100;
    end
endmodule
```

test different scenarios