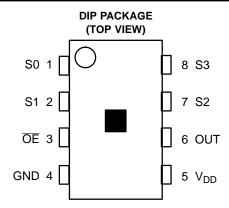


## TSL230R-LF, TSL230AR-LF, TSL230BR-LF PROGRAMMABLE LIGHT-TO-FREQUENCY CONVERTERS

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- High-Resolution Conversion of Light Intensity to Frequency With No External Components
- Programmable Sensitivity and Full-Scale Output Frequency
- Communicates Directly With a Microcontroller
- Single-Supply Operation Down to 2.7 V, With Power-Down Feature
- Absolute Output Frequency Tolerance of ±5% (TSL230BR-LF)
- Nonlinearity Error Typically 0.2% at 100 kHz
- Stable 150 ppm/°C Temperature Coefficient
- Replacements for TSL230, TSL230R, TSL230A, TSL230AR, TSL230B and TSL230BR
- RoHS Compliant

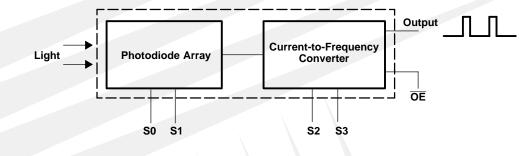


## **Description**

The TSL230R–LF, TSL230AR–LF, and TSL230BR–LF programmable light-to-frequency converters combine a configurable silicon photodiode and a current-to-frequency converter on single monolithic CMOS integrated circuits. The output can be either a pulse train or a square wave (50% duty cycle) with frequency directly proportional to light intensity. Device sensitivity is selectable in three ranges, providing two decades of adjustment. The full-scale output frequency can be scaled by one of four preset values. All inputs and the output are TTL compatible, allowing direct two-way communication with a microcontroller for programming and output interface. An output enable ( $\overline{OE}$ ) is provided that places the output in the high-impedance state for multiple-unit sharing of a microcontroller input line.

The devices are available with absolute-output-frequency tolerances of  $\pm 5\%$  (TSL230BR-LF),  $\pm 10\%$  (TSL230AR-LF), or  $\pm 20\%$  (TSL230R-LF). They have been temperature compensated for the ultraviolet-to-visible light range of 320 nm to 700 nm and respond over the light range of 320 nm to 1050 nm. The devices are characterized over the temperature range of  $-25^{\circ}$ C to  $70^{\circ}$ C.

### **Functional Block Diagram**



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# TSL230R-LF, TSL230AR-LF, TSL230BR-LF PROGRAMMABLE LIGHT-TO-FREQUENCY CONVERTERS

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#### **Terminal Functions**

TERMII	TERMINAL		DECORIDATION						
NAME	NO.	TYPE	DESCRIPTION						
GND	4		Ground						
ŌĒ	3	ı	Enable for f <sub>O</sub> (active low)						
OUT	6	0	Scaled-frequency (f <sub>O</sub> ) output						
S0, S1	1, 2	1	Sensitivity-select inputs						
S2, S3	7, 8	I	f <sub>O</sub> scaling-select inputs						
$V_{DD}$	5		Supply voltage						

## **Selectable Options**

S1	S0	SENSITIVITY					
L	L	Power down					
L	Н	1×					
Н	L	10×					
Н	Н	100×					

S3	S2	f <sub>O</sub> SCALING (divide-by)
L	L	1
L	Н	2
Н	L	10
Н	Н	100

## **Available Options**

DEVICE T <sub>A</sub>		PACKAGE - LEADS	PACKAGE DESIGNATOR	ORDERING NUMBER		
TSL230R-LF	-25°C to 70°	PDIP-8	Р	TSL230R-LFP		
TSL230AR-LF	–25°C to 70°	PDIP-8	Р	TSL230AR-LFP		
TSL230BR-LF	−25°C to 70°	PDIP-8	Р	TSL230BR-LFP		

## Absolute Maximum Ratings over operating free-air temperature range (unless otherwise noted)<sup>†</sup>

Supply voltage, V <sub>DD</sub> (see Note 1)	6 V
Input voltage range, all inputs, V <sub>I</sub>	$\dots$ -0.3 V to V <sub>DD</sub> + 0.3 V
Operating free-air temperature range, T <sub>A</sub>	–25°C to 70°C
Storage temperature range	–25°C to 85°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds <sup>‡</sup>	

<sup>&</sup>lt;sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

NOTE 1: All voltage values are with respect to GND.

## **Recommended Operating Conditions**

		MIN	NOM	MAX	UNIT
Supply voltage, V <sub>DD</sub>	2.7	5	5.5	V	
High-level input voltage, V <sub>IH</sub>	V <sub>DD</sub> = 4.5 V to 5.5 V	2		$V_{DD}$	V
Low-level input voltage, V <sub>IL</sub>	V <sub>DD</sub> = 4.5 V to 5.5 V	0		8.0	V
Operating free-air temperature range, T <sub>A</sub>	-25		70	°C	



<sup>&</sup>lt;sup>‡</sup> Not recommended for solder reflow.

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## Electrical Characteristics at $T_A = 25^{\circ}C$ , $V_{DD} = 5 \text{ V}$ (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V <sub>OH</sub>	High-level output voltage	$I_{OH} = -4 \text{ mA}$	4	4.5		V
$V_{OL}$	Low-level output voltage	I <sub>OL</sub> = 4 mA		0.25	0.4	V
I <sub>IH</sub>	High-level input current				5	μΑ
I <sub>IL</sub>	Low-level input current				5	μΑ
	Ourselv surrent	Power-on mode		2	3	mA
IDD	Supply current	Power-down mode		5	12	μΑ
F.S.	Full-scale frequency <sup>†</sup>		1.1			MHz
	Temperature coefficient of output frequency	λ ≤ 700 nm		±150		ppm/°C
k <sub>SVS</sub>	Supply voltage sensitivity	V <sub>DD</sub> = 5 V ±10%		±0.5		%/V

<sup>&</sup>lt;sup>†</sup> Full-scale frequency is the maximum operating frequency of the device without saturation.

## Operating Characteristics at V<sub>DD</sub> = 5 V, T<sub>A</sub> = 25°C, E<sub>e</sub> = 130 $\mu$ W/cm², $\lambda_p$ = 640 nm (unless otherwise noted)

			TSL230R-LF			TSL230AR-LF			TSL230BR-LF			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNIT
		S0 = S1 = H, S2 = S3 = L	80	100	120	90	100	110	95	100	105	kHz
		S1 = H, S0 = S2 = S3 = L	8	10	12	9	10	11	9.5	10	10.5	kHz
		S0 = H, S1 = S2 = S3 = L	0.8	1	1.2	0.9	1	1.1	0.95	1	1.05	kHz
f <sub>O</sub>	Output frequency	S0 = S1 = S2 = H, S3 = L	40	50	60	45	50	55	47.5	50	52.5	kHz
		S0 = S1 = S3 = H, S2 = L	8	10	12	9	10	11	9.5	10	10.5	kHz
		S0 = S1 = S2 = S3 = H	0.8	1	1.2	0.9	1	1.1	0.95	1	1.05	kHz
		$E_e = 0$ , $S0 = S1 = H$ , $S2 = S3 = L$		0.4	10		0.4	10		0.4	10	Hz
R <sub>e</sub>	Responsivity	S0 = S1 = H, S2 = S3 = L		0.77			0.77			0.77		kHz/ (μW/ cm <sup>2</sup> )
	Output pulse duration	S2 = S3 = L	125		600	125		600	125		600	ns
t <sub>w</sub>		S2 or S3 = H		1/2f <sub>O</sub>			1/2f <sub>O</sub>			1/2f <sub>O</sub>		s
		$f_O = 0$ MHz to 10 kHz		±0.1%			±0.1%			±0.1%		%F.S.
	Nonlinearity ‡#	$f_O = 0$ MHz to 100 kHz		±0.2%			±0.2%			±0.2%		%F.S.
		$f_O = 0$ MHz to 1 MHz		±0.5%			±0.5%			±0.5%		%F.S.
	Recovery from power down				100			100			100	μs
	Step response to full-scale step input				1 pul	se of nev	v frequer	ncy plus	1 μs			
	Response time to programming change			2 p	eriods o	f new pri	ncipal fre	equency	plus 1 μ	s§		
	Response time to output enable (OE)			50	150		50	150		50	150	ns

<sup>&</sup>lt;sup>‡</sup> Nonlinearity is defined as the deviation of f<sub>O</sub> from a straight line between zero and full scale, expressed as a percent of full scale.

<sup>§</sup> Principal frequency is the internal oscillator frequency, equivalent to divide-by-1 output selection.

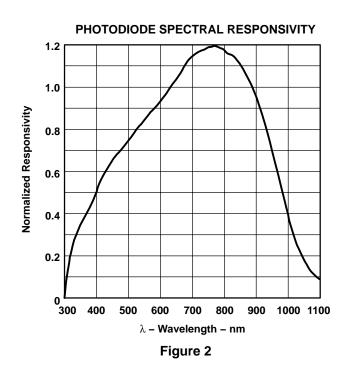


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<sup>#</sup> Nonlinearity test condition: S0 = S1 = H, S2 = S3 = L.

## **TYPICAL CHARACTERISTICS**

## **OUTPUT FREQUENCY IRRADIANCE** 1000 $V_{DD} = 5 V$ $\lambda_p = 640 \text{ nm}$ 100 . T<sub>A</sub> = 25°C S2 = S3 = Lf<sub>O</sub> - Output Frequency - kHz 10 S0 = H, S1 = H0.1 S0 = L, S1 = H0.01 S0 = H, S1 = L0.001 0.001 0.01 0.1 10 100 1 k 10 k 100 k 1 M E<sub>e</sub> - Irradiance - μW/cm<sup>2</sup>



# DARK FREQUENCY

Figure 1

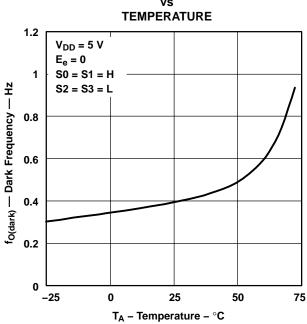
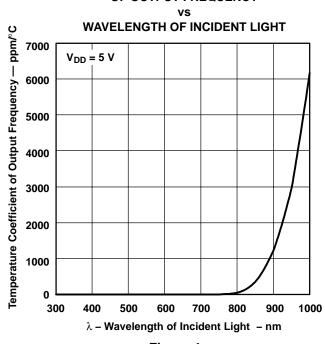


Figure 3

## TEMPERATURE COEFFICIENT OF OUTPUT FREQUENCY



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#### TYPICAL CHARACTERISTICS

## **OUTPUT FREQUENCY SUPPLY VOLTAGE** 1.010 $T_A = 25^{\circ}C$ $f_0 = 100 \text{ kHz}$ 1.005 Normalized Output Frequency 1.000 0.995 0.990 0.985 0.980 2.5 3 5.5 3.5 4.5 5 V<sub>DD</sub> - Supply Voltage - V

### **APPLICATION INFORMATION**

Figure 5

### **Power-supply considerations**

Power-supply lines must be decoupled by a  $0.01-\mu F$  to  $0.1-\mu F$  capacitor with short leads placed close to the device package. A low-noise power supply is required to minimize jitter on output pulses.

## Input interface

A low-impedance electrical connection between the device  $\overline{OE}$  pin and the device GND pin is required for improved noise immunity.

#### **Output interface**

The output of the device is designed to drive a standard TTL or CMOS logic input over short distances. If lines greater than 12 inches are used on the output, a buffer or line driver is recommended.

## Sensitivity adjustment

Sensitivity is controlled by two logic inputs, S0 and S1. Sensitivity is adjusted using an electronic iris technique — effectively an aperture control — to change the response of the device to a given amount of light. The sensitivity can be set to one of three levels:  $1 \times$ ,  $10 \times$  or  $100 \times$ , providing two decades of adjustment. This allows the responsivity of the device to be optimized to a given light level while preserving the full-scale output-frequency range. Changing of sensitivity also changes the effective photodiode area by the same factor.



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#### APPLICATION INFORMATION

## **Output-frequency scaling**

Output-frequency scaling is controlled by two logic inputs, S2 and S3. Scaling is accomplished on chip by internally connecting the pulse-train output of the converter to a series of frequency dividers. Divided outputs available are divide-by 2, 10, 100, and 1 (no division). Divided outputs are 50 percent-duty-cycle square waves while the direct output (divide-by 1) is a fixed-pulse-width pulse train. Because division of the output frequency is accomplished by counting pulses of the principal (divide-by 1) frequency, the final-output period represents an average of n (where n is 2, 10, or 100) periods of the principal frequency. The output-scaling-counter registers are cleared upon the next pulse of the principal frequency after any transition of the S0, S1, S2, S3, or OE lines. The output goes high upon the next subsequent pulse of the principal frequency, beginning a new valid period. This minimizes the time delay between a change on the input lines and the resulting new output period in the divided output modes. In contrast with the sensitivity adjust, use of the divided outputs lowers both the full-scale frequency and the dark frequency by the selected scale factor.

The frequency-scaling function allows the output range to be optimized for a variety of measurement techniques. The divide-by-1 or straight-through output can be used with a frequency counter, pulse accumulator, or high-speed timer (period measurement). The divided-down outputs may be used where only a slower frequency counter is available, such as a low-cost microcontroller, or where period measurement techniques are used. The divide-by-10 and divide-by-100 outputs provide lower frequency ranges for high resolution-period measurement.

### Measuring the frequency

The choice of interface and measurement technique depends on the desired resolution and data acquisition rate. For maximum data-acquisition rate, period-measurement techniques are used.

Using the divide-by-2 output, data can be collected at a rate of twice the output frequency or one data point every microsecond for full-scale output. Period measurement requires the use of a fast reference clock with available resolution directly related to reference-clock rate. Output scaling can be used to increase the resolution for a given clock rate or to maximize resolution as the light input changes. Period measurement is used to measure rapidly varying light levels or to make a very fast measurement of a constant light source.

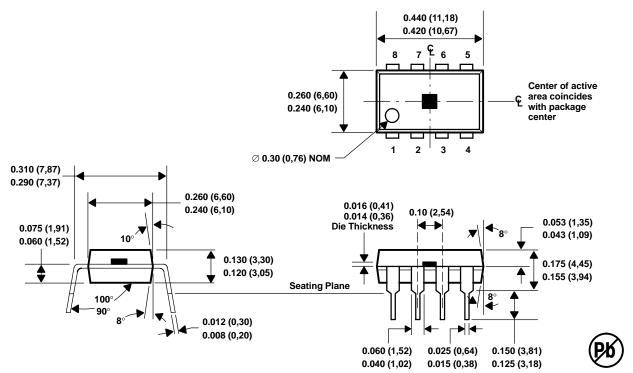
Maximum resolution and accuracy may be obtained using frequency-measurement, pulse-accumulation, or integration techniques. Frequency measurements provide the added benefit of averaging out random or high-frequency variations (jitter) resulting from noise in the light signal or from noise in the power supply. Resolution is limited mainly by available counter registers and allowable measurement time. Frequency measurement is well suited for slowly varying or constant light levels and for reading average light levels over short periods of time. Integration (the accumulation of pulses over a very long period of time) can be used to measure exposure, the amount of light present in an area over a given time period.



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### **MECHANICAL INFORMATION**

This dual-in-line package consists of an integrated circuit mounted on a lead frame and encapsulated with an electrically nonconductive clear plastic compound. The photodiode area is typically  $0.92 \text{ mm}^2$   $(0.0014 \text{ in}^2)$  (S0 = S1 = H).



NOTES: A. All linear dimensions are in inches and (millimeters).

- B. Index of refraction of clear plastic is 1.55.
- C. Lead finish is NiPd.
- D. This drawing is subject to change without notice.

Figure 6. Plastic Dual-In-Line Packaging Configuration

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